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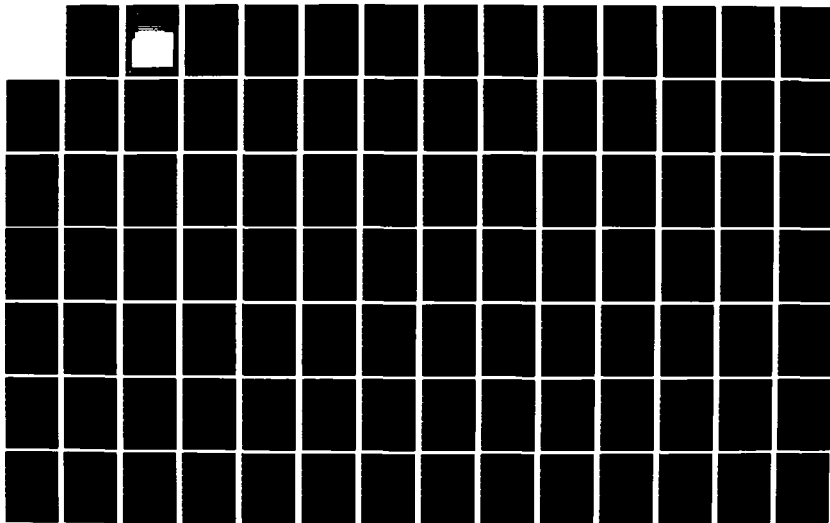
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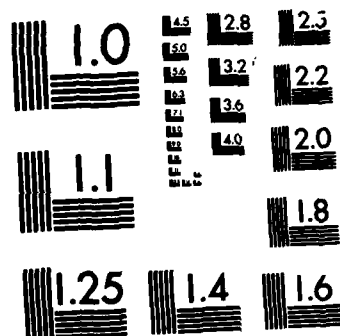
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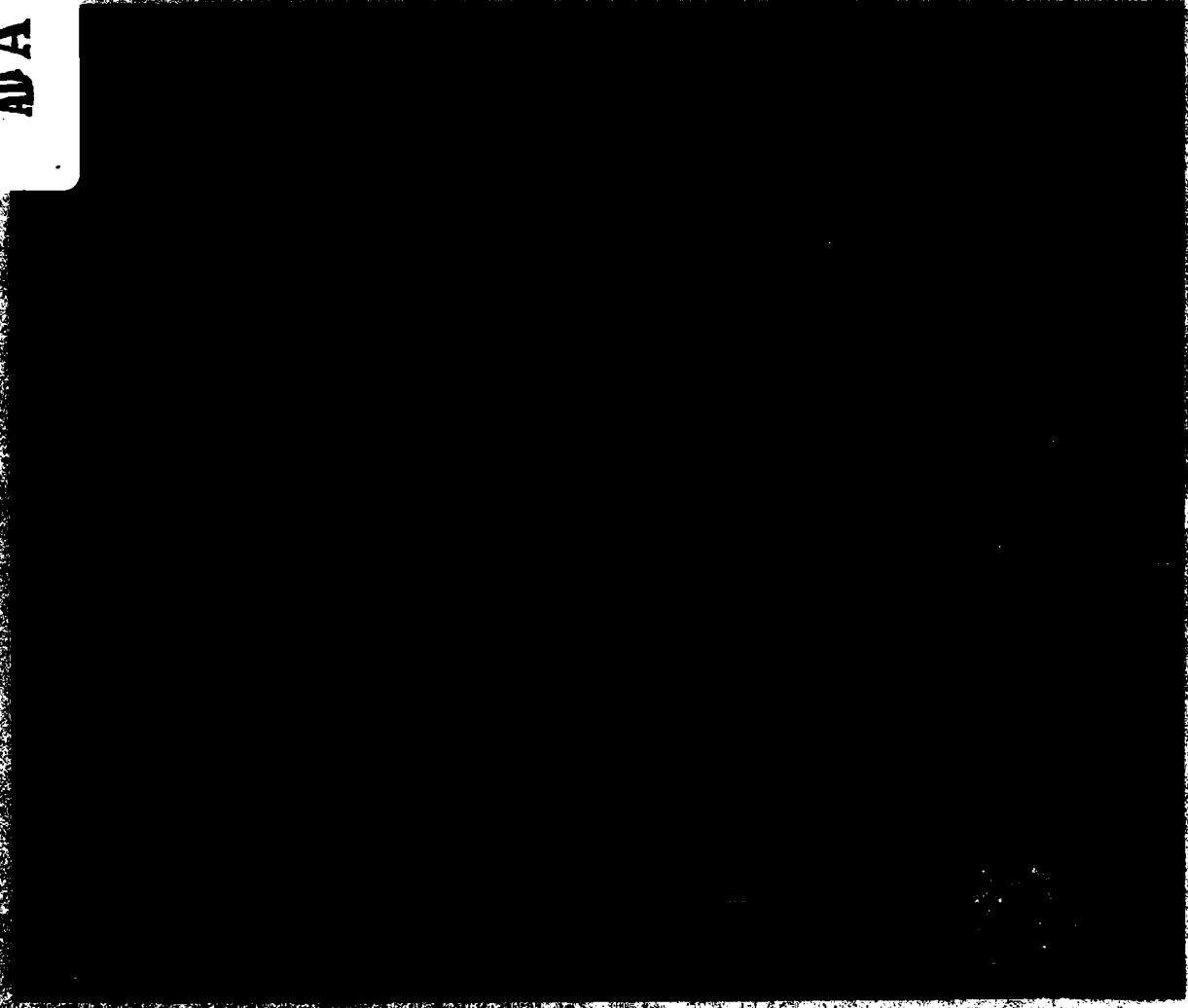
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**NONEXPLOIT CIRCULAR
PERTURBATIONS AND
INTERCOMPARISON STUDIES**



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AND INTERCONNECTED SYSTEMS

by

George Michael Peponides

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NONEXPLICIT SINGULAR PERTURBATIONS
AND INTERCONNECTED SYSTEMS

BY

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Dipl., University of Athens, 1976
M.S., University of Illinois, 1979

THESIS

Submitted in partial fulfillment of the requirements
for the degree of Doctor of Philosophy in Electrical Engineering
in the Graduate College of the
University of Illinois at Urbana-Champaign, 1982

Thesis Adviser: Professor P. V. Kokotovic

Urbana, Illinois

TO MY FATHER, AND IN MEMORY OF MY DEAR MOTHER

NONEXPLICIT SINGULAR PERTURBATIONS
AND INTERCONNECTED SYSTEMS

George Michael Peponides, Ph.D.
Department of Electrical Engineering
University of Illinois at Urbana-Champaign, 1982

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CHAPTER 1

INTRODUCTION

1.1 Model Simplification in Large Systems

An issue of paramount importance in the study of large scale systems is that of model simplification or reduced order modeling. The sheer size on the one hand and the richness and complexity of phenomena on the other make the use of detailed models in the analysis and control of large systems impractical if not impossible. A good analyst or designer knows that a model should encompass only the "relevant" behavior of the system and should not be cluttered with unnecessary detail. Although this may sometimes be accomplished by employing parsimonious models for the components of the system, there are cases where further simplification is needed to make the model manageable both computationally and conceptually. A characteristic example arises in stability studies of interconnected power systems where the use of the crudest model for each generator (the so-called electromechanical model) results in hundreds or even thousands of state variables. It is thus desirable to have systematic model order reduction methods for which the approximation involved can be estimated.

Singular perturbations is a well documented [1-4] method for reduced order analysis and design, in which dynamic phenomena of widely different speeds are treated separately. In the short run the slow dynamics are essentially constant and the focus is on the fast ones. In the long run the fast dynamics settle to their "quasi-steady-state" and the focus is on the slow dynamics. This time-scale thinking is common

in diverse engineering fields [5-7]. If a small parameter ϵ representing the speed ratio of slow and fast dynamics can be identified this intuitively appealing idea leads to asymptotic analysis. Most of the literature [8-10] is devoted to systems of the form

$$\begin{aligned} \dot{y} &= f(y, z, \epsilon) & y(0) &= y_0 \\ \epsilon \dot{z} &= g(y, z, \epsilon) & z(0) &= z_0 \end{aligned} \quad (1.1)$$

where ϵ multiplies the z -derivatives, y is a ν -vector and z is a ρ -vector. Formally setting $\epsilon=0$ in (1.1), solving

$$0 = g(\bar{y}, \bar{z}, 0) \quad (1.2)$$

for \bar{z} , $\bar{z} = \Psi(\bar{y})$ and substituting into (1.1)

$$\frac{d\bar{y}}{dt} = f(\bar{y}, \Psi(\bar{y}), 0) \triangleq \bar{f}(\bar{y}), \quad \bar{y}(0) = y_0 \quad (1.3)$$

we obtain the slow reduced model. Writing system (1.1) in the "stretched" time variable $\tau = \frac{t}{\epsilon}$ and setting $\epsilon=0$ we obtain the fast reduced system (or associated system or boundary layer system)

$$\frac{d\tilde{z}}{d\tau} = g(y_0, \bar{z}_0 + \tilde{z}, 0) \quad \tilde{z}(0) = z_0 - \bar{z}_0 \quad (1.4)$$

where $z_0 = \bar{z}(0)$. Variables y, z, ϵ, t are restricted to lie in a domain

$D : \|y - \bar{y}(t)\| < r, \|z - \bar{z}(t)\| < r, 0 \leq \epsilon \leq \epsilon_0, 0 \leq t \leq T$, where $r > \|y_0 - \bar{y}(0)\|$.

The following theorem relates the solutions of (1.1) with the solutions of (1.3)-(1.4).

Theorem 1.1 [35] Let the following conditions be satisfied.

- H1. $f, \partial f / \partial y, \partial f / \partial z, g, \partial g / \partial y, \partial g / \partial z$ are of class C^0 in D .
- H2. The solution $\tilde{z}(\tau)$ of (1.4) exists on $\tau \in [0, \infty]$, is unique, and is asymptotically stable with respect to $\tilde{z}=0$.
- H3. The solution $\bar{y}(t)$ of the reduced system (1.3) exists and is unique on $t \in [0, T]$.
- H4. The real parts of the eigenvalues of the Jacobian matrix

$$\partial g / \partial z(\bar{y}, \bar{z}, 0) \quad (1.5)$$

are negative on $[0, T]$, for $\bar{z} = \Psi(y)$.

Then for sufficiently small ε , the full system (1.1) has a unique solution $y(t, \varepsilon)$ on $t \in [0, T]$ satisfying the initial condition $y(0, \varepsilon) = y_0$, $z(0, \varepsilon) = z_0$. Furthermore,

$$\lim_{\varepsilon \rightarrow 0} y(t, \varepsilon) = \bar{y}(t), \quad \text{on } [0, T] \quad (1.6)$$

$$\lim_{\varepsilon \rightarrow 0} z(t, \varepsilon) = \bar{z}(t) + \tilde{z}\left(\frac{t}{\varepsilon}\right) \quad \text{on } [0, T] \quad (1.7)$$

where the limits in (1.6), (1.7) are uniform in t on $[0, T]$.

From (1.6), (1.7) the response of y in (1.1) is approximated, to $O(\varepsilon)$, by the response of the slow system (1.3) whereas the response of z is approximated by a boundary layer $\tilde{z}(\frac{t}{\varepsilon})$ superimposed on the quasi-steady-state $\bar{z}(t) = \Psi(\bar{y}(t))$.

An extensive literature dealing with system (1.1) and the corresponding controlled system includes results on stability [20, 65, 66], linear [4, 67] and nonlinear [3] regulator design, controllability properties [37] and time-optimal control [68], filtering and smoothing [69]. A basic

assumption in these references is that the Jacobian matrix $\partial g / \partial z(\bar{y}, \bar{z}, 0)$ is nonsingular. When this happens we say that time scales in (1.1) are explicit, that is, they coincide with the decomposition of the state vector into y and z . When, however, $\partial g / \partial z$ is singular time scales in (1.1) are nonexplicit, that is, all states may be mixed having fast and slow parts. Some authors treat such cases as "singular-singularly perturbed" systems [11-13] or "generalized singularly perturbed" systems [14,15].

In this thesis we take an alternate route. We recognize that in a wide class of systems singularity of $\partial g / \partial z$ is due to the selection of state variables; hence a nonsingular transformation removes the singularity of $\partial g / \partial z$, and defines new states in which time-scales are explicit. This approach has two advantages. First it puts the system into a form in which the results alluded to before can be applied. Second, from the transformed system we can easily define fast and slow reduced systems describing the system behavior in the short run and in the long run. Following this approach we establish a relation between weak connections and time scales in a class of interconnected systems. Separation of time scales in such systems leads to a physical decomposition into a slow core and a number of weakly coupled fast subsystems. The results are further specialized to structured interconnected systems such as power systems and other dynamic networks.

1.2 Chapter Preview

Chapter 2 starts with a simple RC example pointing the relationship between equilibrium and conservation properties on the one hand and time scales on the other [16]. These properties are used in the construction of

a transformation that makes the time scales in Linear Time Invariant (LTI) systems explicit. Next a multi-time-scale system is viewed as a succession of two-time-scale ones. Starting from the fastest time scale we proceed to the slower ones, using at each step the transformation that makes time-scales explicit. This procedure defines a sequence of "nested" reduced order models. The transformation separating the time scales is then generalized to LTI systems with inputs.

In Chapter 3 the equilibrium and conservation reasoning is extended to nonlinear systems leading to a transformation that makes time scales explicit in models of the form

$$\epsilon \dot{x} = h(x, \epsilon). \quad (1.8)$$

It is then shown that the results of [17-19] on high gain feedback and disturbance decoupling generalize to a class of nonlinear systems for which the controls enter linearly but the output map and feedback law are nonlinear. We next turn to interconnected systems made of systems with equilibrium manifolds and show that weak connections give rise to two-time-scale behavior. A decomposition of interconnected systems into a slow core and fast local systems leads to decentralized stability criteria based on the results of [20].

Chapter 4 deals with time scales, coherency and aggregation in nonlinear dynamic networks. Coherency based aggregation [21-23], a common procedure for order reduction in power systems, is given theoretical foundations for nonlinear electromechanical models, thus extending the results of [24-29]. It is shown that linear physical laws result in linear time-scale separating transformation even when some components of

the network are nonlinear. A five-machine power system example illustrates the proposed reduced-order modeling and verifies its validity.

Extensions in several directions and possible uses of the decomposition in direct stability analysis are discussed in Chapter 5.

CHAPTER 2

MODELING OF TWO-TIME-SCALE SYSTEMS

2.1 Introduction

When the model of a real system with the two-time-scale property is expressed in terms of physical variables it often fails to be in the form (1.1). An important requirement in (1.1) is that $\partial g / \partial z$ be nonsingular along $\bar{z}(t)$. When this condition is violated the model is said to be non-explicit and the conclusions in Chapter 1 have to be modified.

Some authors [11-13] treat nonexplicit models as "singularly perturbed" systems. Instead we approach them from the modeling point of view recognizing that the singularity of $\partial g / \partial z$ is due to the choice of state variables. We show that equilibrium [16] and conservation properties provide a coordinate-free characterization of singular perturbations. These properties are used in the construction of a transformation leading to the explicit model (1.1) with the slow part of $z(t)$ being $O(\epsilon)$. The discussion in this chapter is restricted to Linear Time Invariant (LTI) systems. Extension of the basic ideas to nonlinear systems and applications to interconnected systems, high gain feedback and dynamic networks appear in Chapters 3 and 4.

In Section 2.2, a simple physical system is used to motivate the discussion and indicate the relation between time scales on the one hand and equilibrium and conservation properties on the other. In Section 2.3, the relation is established for LTI systems and a transformation is constructed that transforms a nonexplicit singularly perturbed model to the explicit

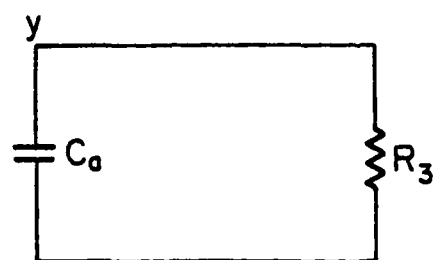
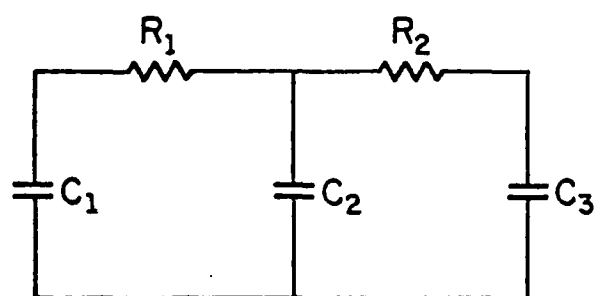
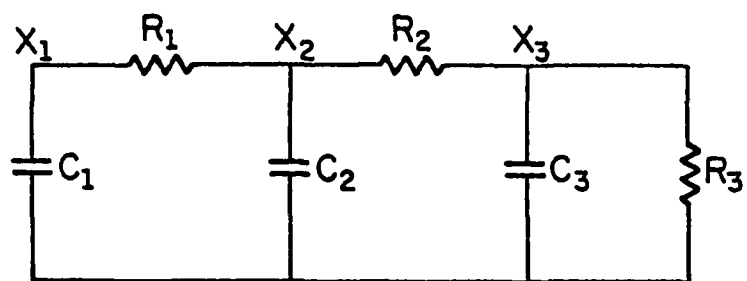
model (1.1). In Section 2.4, multi-time-scale systems are treated as a succession of two-time-scale systems and a sequence of nested reduced order models is defined. Section 2.5 generalizes the results of Section 2.3 to systems with inputs and Section 2.6 deals with some structured nonexplicit models.

2.2 Equilibrium and Conservation Properties

Although nonexplicit singular perturbations occur in as simple systems as RC-circuits they have not attracted much attention. In contrast, explicit perturbations have been investigated for networks with "parasitic" inductances and capacitances [30,31]. When such parasitics are expressed as multiples of ϵ and capacitor voltages and inductor currents are used as state variables, the circuit model is in the explicit form (1.1). A simple illustration is the RC-circuit of Fig. 2.1a with state equations

$$\begin{aligned} (R_1 C_1) \, dx_1/dt_d &= -x_1 + x_2 \\ (R_1 C_2) \, dx_2/dt_d &= x_1 - [1 + (R_1/R_2)] x_2 + (R_1/R_2) x_3 \\ (R_1 C_3) \, dx_3/dt_d &= (R_1/R_2) x_2 - [(R_1/R_2) + (R_1/R_3)] x_3 \end{aligned} \quad (2.1)$$

where the capacitor voltages were chosen as states and t_d is dimensional time. Suppose that C_2 and C_3 are "parasitic," say $C_2=C_3=\epsilon C_1$ and that all the resistors are of the same order of magnitude. Recognizing $R_1 C_1$ as a typical large time constant and defining the slow dimensionless time $t=t_d/(R_1 C_1)$ [32,33], (2.1) becomes



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Fig. 2.1 (a) Circuit with R_3 much larger than R_1 , R_2 .
 (b) Fast circuit described by (2.6).
 (c) Slow reduced circuit for y in (2.33).

$$dx_1/dt = -x_1 + x_2$$

$$\epsilon dx_2/dt = x_1 - [1 + (R_1/R_2)] x_2 + (R_1/R_2) x_3 \quad (2.2)$$

$$\epsilon dx_3/dt = (R_1/R_2) x_2 - [(R_1/R_2) + (R_1/R_3)] x_3$$

with ϵ multiplying the derivatives of x_2 and x_3 . Thus x_1 appears as the y-variable and x_2, x_3 as the z-variables of (1.1) and the model is explicit because the two-by-two matrix of x_2, x_3 is nonsingular. The slow reduced model (1.2) represents the circuit with parasitic capacitors C_2 and C_3 opened (Fig. 2.1b), whereas in the fast reduced model (1.3) the large capacitor C_3 is shortened [30] (Fig. 2.1c).

In the same circuit nonexplicit singular perturbations occur when all capacitors are of the same order of magnitude, say $C_1=C_2=C_3=C$, but the resistors are not. For example, if R_1 and R_2 are small and R_3 is large, say

$$R_1 = r, R_2 = r/2, R_3 = R, r/R = \epsilon \quad (2.3)$$

typical large and small time constants are RC and rC , respectively, and in the dimensionless time variables

$$t = \frac{t_d}{RC}, \tau = \frac{t_d}{rC}, \frac{t}{\tau} = \epsilon \quad (2.4)$$

the circuit is described by

$$\epsilon (dx/dt) = dx/d\tau = \begin{bmatrix} -1 & 1 & 0 \\ 1 & -3 & 2 \\ 0 & 2 & -2-\epsilon \end{bmatrix} x \triangleq A(\epsilon) x. \quad (2.5)$$

Note that ϵ multiplies all the derivatives in the slow time-scale t and thus there are no explicit slow y -variables in the system. If $\frac{\partial g}{\partial z}$, that is $A(0)$, were nonsingular, no slow phenomenon would exist in (2.5) and the system would not possess the two-time-scale property. However, $A(0)$ is singular indicating the existence of a "hidden" slow phenomenon. To see this assume that $x(0) = [1 \ 1 \ 1]^T$ in (2.5). Then the slow-time derivatives dx/dt remain finite when $\epsilon \rightarrow 0$ suggesting that (2.5) is a two-time-scale system. Physically the slow phenomenon is the discharge of the capacitors through the large "leakage" resistor R_3 . Neglecting this "leakage," Fig. 2.1b, makes the slow phenomenon infinitely slow, that is, constant and corresponds to setting $\epsilon=0$ in the τ -model of (2.5)

$$\frac{dx}{d\tau} = \begin{bmatrix} -1 & 1 & 0 \\ 1 & -3 & 2 \\ 0 & 2 & -2 \end{bmatrix} x = A(0)x. \quad (2.6)$$

Since $A(0)$ is singular the equation

$$A(0) x = 0 \quad (2.7)$$

has an infinite number of solutions given by

$$x = \alpha [1 \ 1 \ 1]^T \quad (2.8)$$

where α is any real number, that is, (2.6) has a continuum of equilibrium points. This can be seen from the circuit of Fig. 2.1b where any x such that

$$x_1 - x_2 = 0, x_3 - x_2 = 0 \quad (2.9)$$

is an equilibrium point. The line represented by (2.8)-(2.9) will be denoted by S.

Kirchhoff's current law (KCL) applied to the ground node of Fig. 2.1b gives the dual property that is the conservation of total charge for all τ ,

$$C_1 x_1(\tau) + C_2 x_2(\tau) + C_3 x_3(\tau) = C_1 x_1(0) + C_2 x_2(0) + C_3 x_3(0) \quad (2.10)$$

which means that every trajectory $x(\tau)$ of (2.6) is confined to a plane F passing through the initial point $x(0)$ orthogonal to the vector $[C_1 \ C_2 \ C_3]^T$. The quantity in (2.10), constant when $\epsilon=0$, becomes slowly varying when $\epsilon > 0$, that is when the "leakage" R_3 is introduced. A circuit describing this slow phenomenon is given in Fig. 2.1c and will be derived in the next section.

From the above discussion we conclude that the trajectories $x(t)$ of the original system 2.5 consist of two distinct parts. First in a "boundary layer" near plane F the state $x(t)$ rapidly approaches line S. Then, from a neighborhood of the intersection of plane F with line S, $x(t)$ continues to slowly "slide" along line S. The geometry of this situation is sketched in Fig. 2.2. Note that the behavior of $x(t)$ is similar to that of the explicit model, that is, a fast transient is followed by a slow motion close to a line of "quasi-equilibria" S. The basic difference is that in the explicit model (2.2) the plane F close to which the boundary layer occurs, is orthogonal to axis x_1 (Fig. 2.3).

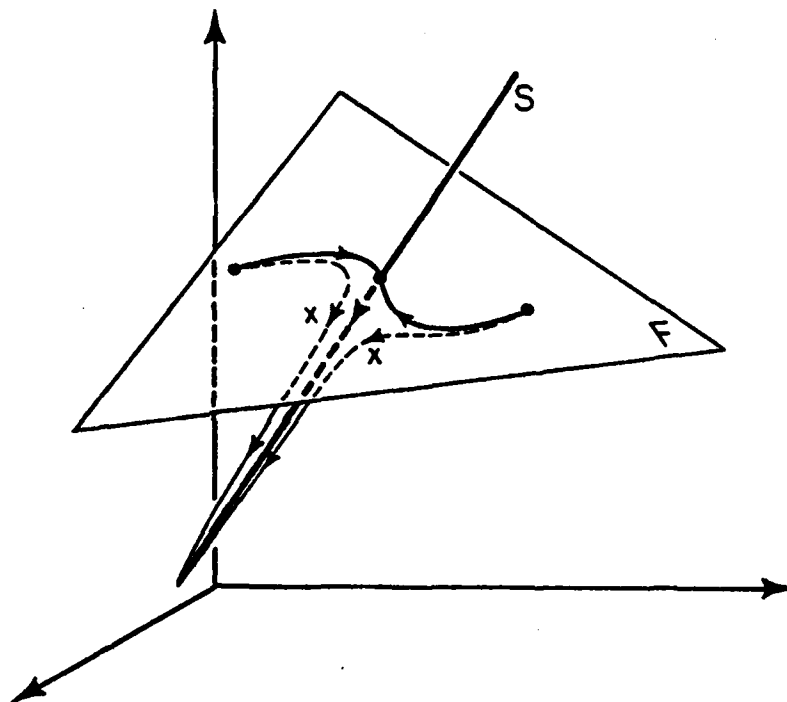
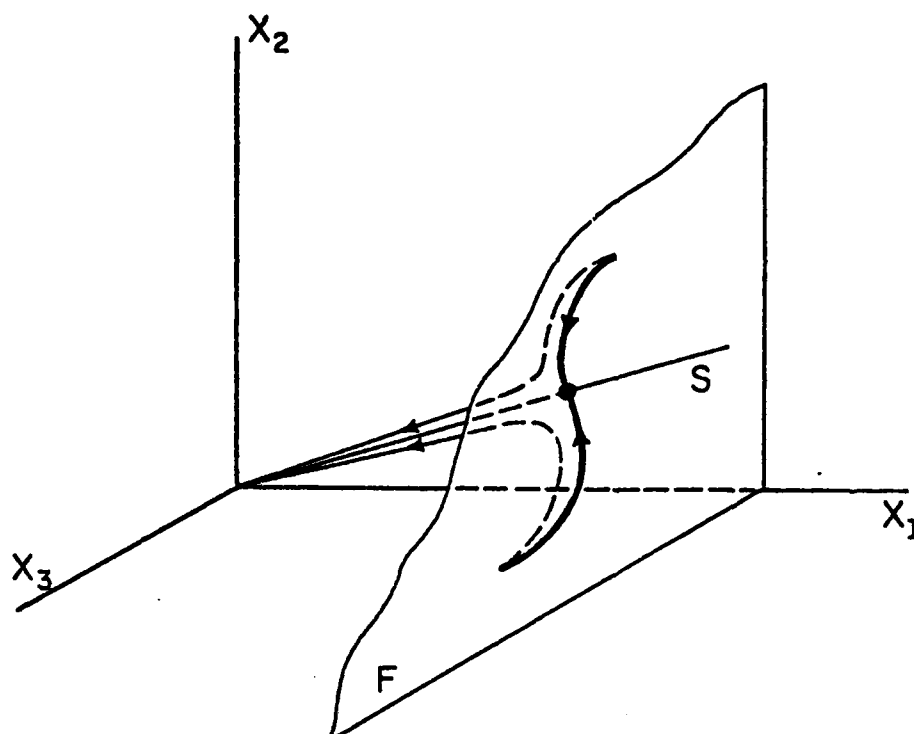


Fig. 2.2 Trajectories of the auxiliary circuit in Fig. 2.1b lie on F. Trajectories of the actual circuit Fig. 2.1a, are denoted by x.



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Fig. 2.3 Equilibrium (S) and dynamic manifolds (F) of the explicit model 2.2.

Examination of Fig. (2.2)-(2.3) indicates that in nonexplicit models fast dynamics are observed by all states whereas in explicit models they are only weakly observed by some states (the y-variables).

This example indicates that the time scales of the original system (2.5) are related to the equilibrium and conservation properties of the auxiliary system (2.6) in τ -scale. These properties are coordinate free and characterize all two-time-scale systems reducible to the explicit model. In the next section they will serve for a choice of coordinates in which the time scales are explicit.

2.3 Nonexplicit Singularly Perturbed Systems

The discussion of the previous section will now be generalized to the system

$$\varepsilon dx/dt = dx/d\tau = A(\varepsilon) x \quad (2.11)$$

where $x \in \mathbb{R}^n$, $A(\varepsilon)$ is a time invariant $n \times n$ matrix depending on ε , and t , τ the slow and fast time variables, respectively. The following is assumed about $A(\varepsilon)$.

Assumption 2.1 $A(\varepsilon)$ can be written as

$$A(\varepsilon) = A_0 + \varepsilon A_1(\varepsilon) \quad (2.12)$$

with $A_1(\varepsilon)$ bounded at $\varepsilon=0$ and $A(0)=A_0$, satisfying

$$\mathcal{R}(A_0) \oplus \mathcal{N}(A_0) = \mathbb{R}^n \quad (*) \quad (2.13)$$

(*) In the terminology of [13], (2.13)-(2.14) is equivalent to $\text{ind } A_0 = 1$. In [13] it is shown that this condition is necessary for $\lim_{\varepsilon \rightarrow 0} x(t)$ to exist.

where $R(A_0)$ is the range space of A_0 , $\eta(A_0)$ is the null space of A_0 , and \oplus denotes the direct sum of two spaces [34]. The dimensions of $R(A_0)$, $\eta(A_0)$ are

$$\dim R(A_0) = \rho \geq 1, \dim \eta(A_0) = \nu \geq 1, \rho + \nu = n. \quad (2.14)$$

Equation (2.13) is equivalent to saying that A_0 has a complete set of eigenvectors corresponding to its zero eigenvalues, which in turn is equivalent to the following: $R(A_0)$ is the invariant space (eigenspace) of A_0 corresponding to the nonzero eigenvalues, and $\eta(A_0)$ is the invariant space (eigenspace) corresponding to the zero eigenvalues.

To study the time-scale behavior of (2.11) the auxiliary system

$$dx/d\tau = A_0 x \quad (2.15)$$

is defined with A_0 as in (2.12). By assumption, (2.15) has a ν -dimensional equilibrium manifold^(*) (i.e. $\eta(A_0)$) S consisting of all x such that

$$A_0 x = 0. \quad (2.16)$$

If W is a $\rho \times n$ matrix, $\text{rank } W = \rho$, whose rows span the row space of A , then [34]

$$Wx = 0 \quad \forall x \in S \quad (2.17)$$

To see the conservation property of (2.15) we note that if V is a $\nu \times n$ matrix, $\text{rank } V = \nu$ whose rows span the left null space of A_0 , i.e. $VA_0 = 0$, then

(*) Although S is presently, simply a subspace, we call it manifold in anticipation of the nonlinear extension in Chapter 3.

$$V(dx/d\tau) = V A_0 x = 0 \quad \forall \tau, \forall x(0) \in \mathbb{R}^n. \quad (2.18)$$

Thus, the v -dimensional quantity Vx is constant along the trajectories of (2.15),

$$V x(\tau) = V x(0), \quad \forall x(0) \in \mathbb{R}^n. \quad (2.19)$$

This means that for each value of $V x(0)$ the trajectory of (2.15) is confined to a linear manifold^(*) defined by (2.19).

This linear manifold called dynamic manifold F , is orthogonal to the rows of V and contains the initial point $x(0)$. The orthogonality between the left null space and the range space of a matrix [34] implies that F is a translate of $\mathcal{R}(A_0)$.

The above discussion has established equilibrium (Eq. (2.17)) and conservation (Eq. (2.19)) properties analogous to the ones of the RC-circuit of the previous section (Eq. (2.9) and (2.10)). The behavior of the trajectories is still the one depicted in Fig. 2.2 with S and F defined by (2.17) and (2.19), respectively. We are now ready to define a new set of coordinates in which the time scales are explicit.

Theorem 2.2 Under Assumption (2.1) the change of coordinates

$$y = Vx, \quad z = Wx \quad (2.20)$$

transforms (2.11) into the explicit model (1.1) with $\bar{z}(t) = 0$.

(*) A linear manifold of dimension r is a translation of an r -dimensional subspace.

Proof: Since the rows of V, W form bases for the left null and row spaces of A_0 , respectively, the transformation

$$T = \begin{bmatrix} V \\ W \end{bmatrix} \quad (2.21)$$

defined by (2.20) has inverse

$$T^{-1} = [P \ Q] \quad (2.22)$$

where the columns of P, Q form bases of $\mathcal{N}(A_0)$, $\mathcal{R}(A_0)$. Hence,

$$\begin{aligned} T \left(\frac{A_0}{\epsilon} + A_1(\epsilon) \right) T^{-1} &= \frac{1}{\epsilon} \begin{bmatrix} V A_0 P & V A_0 Q \\ W A_0 P & W A_0 Q \end{bmatrix} + \begin{bmatrix} V A_1(\epsilon) P & V A_1(\epsilon) Q \\ W A_1(\epsilon) P & W A_1(\epsilon) Q \end{bmatrix} \\ &= \begin{bmatrix} V A_1(\epsilon) P & V A_1(\epsilon) Q \\ W A_1(\epsilon) P & \frac{W A_0 Q}{\epsilon} + W A_1(\epsilon) Q \end{bmatrix} \end{aligned} \quad (2.23)$$

$$dy/dt = A_{11}(\epsilon) y + A_{12}(\epsilon) z \quad (2.24)$$

$$\epsilon dz/dt = \epsilon A_{21}(\epsilon) y + A_{22}(\epsilon) z$$

where $A_{11}(\epsilon) \triangleq V A_1(\epsilon) P$, $A_{12}(\epsilon) \triangleq V A_1(\epsilon) Q$, $A_{21}(\epsilon) \triangleq W A_1(\epsilon) P$ and

$$A_{22}(\epsilon) \triangleq W A_0 Q + \epsilon W A_1(\epsilon) Q. \quad (2.25)$$

To show that (2.24) is explicit model we need to show that $A_{22}(0)$ is nonsingular. Notice that Assumption (2.1) implies that $\mathcal{R}(A_0)$ is the

eigenspace of the nonzero eigenvalues of A_0 . Hence there is a $p \times p$ nonsingular matrix G whose eigenvalues are the nonzero eigenvalues of A_0 such that

$$A_0 Q = QG. \quad (2.26)$$

The last relation implies

$$A_{22}(0) = W A_0 Q = W Q G = G \quad (2.27)$$

which is nonsingular.

Remark: Writing

$$x = T^{-1} \begin{bmatrix} y \\ z \end{bmatrix} = Py + Qz \quad (2.28)$$

we see that y, z are the representations of x with respect to bases P, Q of $\mathcal{N}(A_0)$, $\mathcal{R}(A_0)$, respectively. Another way to view (2.28) is that Py is the projection of x on $\mathcal{N}(A_0)$ along $\mathcal{R}(A_0)$ and hence y is the representation of this projection with respect to basis P . A similar interpretation holds for Qz and z .

We now illustrate the application of Theorem 2.2 with the RC-circuit of Fig. 2.1 in which the time scales are due to large and small resistors as in (2.3). The auxiliary system is given in (2.6) and the equilibrium and dynamic manifolds are defined by (2.9) and (2.10), respectively. A choice of coordinates according to these equations is

$$y = (C_1 x_1 + C_2 x_2 + C_3 x_3) / C_a \quad (2.29)$$

$$z_1 = x_1 - x_2, \quad z_2 = x_3 - x_2 \quad (2.30)$$

where the division by

$$C_a = C_1 + C_2 + C_3 \quad (2.31)$$

in (2.29) retains the physical meaning of y as a voltage variable. In the new coordinates the circuit is described by

$$\begin{aligned} dy/dt &= - (C/C_a) y + (C^2/C_a^2) z_1 - (2C^2/C_a^2) z_2 \\ \epsilon(dz_1/dt) &= - 2z_1 - 2z_2 \\ \epsilon(dz_2/dt) &= - \epsilon y - (1 - \epsilon(C/C_a)) z_1 - (4 + \epsilon(2C/C_a)) z_2. \end{aligned} \quad (2.32)$$

As stated in Theorem 2.2 the z -equations in (2.32) give $\bar{z}(t) = 0$. Hence the slow reduced model is

$$d\bar{y}/dt = - (C/C_a)\bar{y} \quad (2.33)$$

represented by the circuit in Fig. 2.1c and the fast reduced model is

$$\begin{aligned} d\tilde{z}_1/d\tau &= - 2\tilde{z}_1 - 2\tilde{z}_2 \\ d\tilde{z}_2/d\tau &= - \tilde{z}_1 - 4\tilde{z}_2 \end{aligned} \quad (2.34)$$

represented by the circuit in Fig. 2.1b where the voltages with respect to the "reference" node 2 are used as states.

It is interesting to note the physical interpretation of the new variables. The slow variable y is proportional to the sum of the charges on the three capacitors and can be considered the voltage on the

"aggregate" capacitor C_a (Eq. (2.31), Fig. 2.1c). Application of Kirchhoff's current law to the ground node of Fig. 2.1a shows that the time derivative of y is proportional to the current in R_3 . Since R_3 is large dy/dt is small and y qualifies as a slow variable. The fast variable z_1 equals the voltage across R_1 which due to the smallness of R_1 diminishes quickly to values close to zero. Hence z_1 qualifies as a fast variable. A similar interpretation holds for z_2 . In Chapter 4 we show that this selection of slow and fast variables is good for a wide class of nonlinear dynamic networks.

2.4 Nested Reduced Order Models

In the previous section we showed how equilibrium and conservation properties are used to transform a nonexplicit singularly perturbed model to an explicit one. Writing the explicit model (2.24) in the fast time τ and letting $\epsilon \rightarrow 0$ we obtain the fast reduced model

$$\frac{d\tilde{z}}{d\tau} = A_{22}(0)\tilde{z} \quad \tilde{z}(0) = z(0). \quad (2.35)$$

Similarly writing the model in the slow time t and letting $\epsilon \rightarrow 0$ we obtain the slow reduced model

$$\frac{d\bar{y}}{dt} = A_{11}(0)\bar{y} \quad \bar{y}(0) = y(0). \quad (2.36)$$

If the eigenvalues of $A_{22}(0)$ have negative real parts the responses of (2.35)-(2.36) are $O(\epsilon)$ approximations of the response of (2.24) over bounded intervals [8-10]. If, in addition, the eigenvalues of the slow system matrix $A_{11}(0)$ have negative real parts the approximation is valid over unbounded intervals [36].

It may happen, however, that some eigenvalues of $A_{11}(0)$ are zero. In these cases the approximation is not valid over unbounded intervals since the response of (2.36) tends to a nonzero constant whereas the response of (2.24) tends to zero. Treatment of (2.24) as a two-time-scale system is inadequate because the system may have more than two time scales. Simple expansions of the eigenvalues and eigenvectors, for example, show that if A_{11} is singular the matrix

$$\begin{bmatrix} \epsilon A_{11} & \epsilon A_{12} \\ \epsilon A_{21} & A_{22} \end{bmatrix}$$

which is the matrix of (2.24) in the fast time scale with A_{11} , A_{12} , A_{21} , A_{22} , independent of ϵ , has $O(\epsilon^2)$ eigenvalues in addition to $O(1)$ and $O(\epsilon)$ ones.

Instead of treating the original system (2.11) as a multi-time-scale one, we prefer to deal with only two time scales at a time. That is, starting from the fastest time $t_1 \triangleq \tau$ we consider the system operating in scale t_1 and $t_2 = \epsilon t_1$ only. Viewed from t_1 speeds are $O(1)$ and the rest are $o(1)$; we do not specify whether they are $O(\epsilon)$, $O(\epsilon^2)$, etc. Changing time scales to the slower t_2 , some speeds are $o(\frac{1}{\epsilon})$ and, in time-scale t_2 , are assumed to reach their quasi-steady-state instantaneously. The rest of the system is again treated as two-time-scale with t_2 the fast time.

To make this idea precise we employ the block diagonalizing transformation [37,38]

$$\begin{bmatrix} \xi \\ \eta \end{bmatrix} = \begin{bmatrix} I - \epsilon^2 HL & - \epsilon H \\ \epsilon L & I \end{bmatrix} \begin{bmatrix} y \\ z \end{bmatrix} \quad (2.37)$$

where L, H satisfy

$$A_{21}(\epsilon) + \epsilon L A_{11}(\epsilon) - A_{22}(\epsilon) L - \epsilon^2 L A_{12}(\epsilon) L = 0 \quad (2.38)$$

and

$$A_{12}(\epsilon) - H (A_{22}(\epsilon) + \epsilon^2 L A_{12}(\epsilon)) + \epsilon (A_{11}(\epsilon) - \epsilon A_{12}(\epsilon) L) = 0. \quad (2.39)$$

In the ξ, η coordinated (2.24) becomes

$$\frac{d\xi}{dt} = (A_{11}(\epsilon) - \epsilon A_{12}(\epsilon) L) \xi \quad (2.40)$$

$$\epsilon \frac{d\eta}{dt} = (A_{22}(\epsilon) + \epsilon^2 L A_{12}(\epsilon)) \eta. \quad (2.41)$$

Since $A_{22}(0)$ is nonsingular (2.35) is a regular perturbation of (2.41) and can be used as a fast reduced order model. However, if $A_{11}(0)$ is singular satisfying Assumption 2.1, (2.40) is nonexplicit singularly perturbed model in the form (2.11). Hence, arguing as in Section 2.3, we apply transformation (2.20) to define slow and fast variables in time scale t_2 . The process can be repeated until all time scales are "peeled off" defining nested reduced order models.

2.5 Systems With Inputs

Although some control design work has been done for generalized singularly perturbed systems [13,14], most of the literature [2-4,39-40] deals with the explicit model

$$\begin{aligned}\dot{y} &= A_{11} y + A_{12} z + B_1 u \\ \epsilon \dot{z} &= A_{21} y + A_{22} z + B_2 u\end{aligned}\tag{2.42}$$

where A_{22} is nonsingular. Note that in the y -equations the gain of the control u is $O(1)$ whereas in the z -equations it is $O(\frac{1}{\epsilon})$. We are interested in conditions under which the nonexplicit model

$$\epsilon \frac{dx}{dt} = \frac{dx}{d\tau} = [A_0 + \epsilon A_1(\epsilon)]x + [B_0 + \epsilon \bar{B}(\epsilon)]u\tag{2.43}$$

where A_0 satisfies Assumption 2.1 and $\bar{B}(\epsilon)$ is differentiable at $\epsilon=0$, can be transformed to the explicit model (2.42).

Assumption 2.3 Let V be a $\nu \times n$ matrix that spans the left null space of A_0 . Then

$$V B_0 = 0\tag{2.44}$$

that is, V is in the left null space of B_0 .

Corollary 2.4 Under Assumptions 2.1, 2.3 the transformation

$$y = Vx, \quad z = Wx\tag{2.45}$$

transforms (2.43) into the explicit model (2.42) with $\bar{z} = -A_{22}^{-1}(0)B_2(0)\bar{u}$ where $B_2(\epsilon) = WB_0 + \epsilon W\bar{B}(\epsilon)$ and \bar{u} is the slow control.

Proof: Follows directly from Theorem 2.2 and (2.44).

Condition (2.44) essentially requires that the control driving the slow variable y have $O(1)$ gain, as in the explicit model (2.42). If this condition is not met the variable y , slow in the free system (2.11), is subjected to high gain control altering the time-scale behavior of the system. We will have more to say about high gain feedback in the next chapter. Condition (2.44) is likely to be satisfied in well defined physical problems as demonstrated by the following example.

Consider the transformer of Fig. 2.4 where L_1 , L_2 are the self-inductances of the coils, M is the mutual inductance and x_1 , x_2 are the currents through the coils. Using x_1 , x_2 as states, the state description of the system is

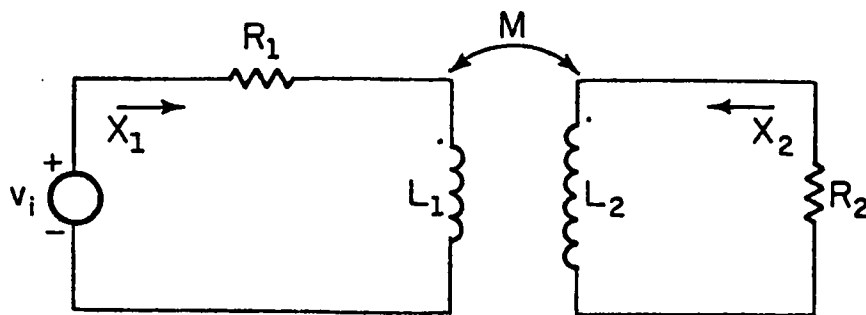
$$\begin{aligned} dx_1/dt &= - (R_1 L_2/d)x_1 + (MR_2/d)x_2 - (L_2/d) v_i \\ dx_2/dt &= (MR_1/d)x_1 - (R_2 L_1/d)x_2 + (M/d) v_i \end{aligned} \quad (2.46)$$

where $d = L_1 L_2 - M^2$.

In the case of an ideal transformer, $d = L_1 L_2 - M^2 = 0$. For nonideal transformer with small leakage

$$\frac{L_1 L_2 - M^2}{L_1 L_2} = \epsilon \quad (2.47)$$

where ϵ is a small positive parameter. Using (2.47) the system matrix of (2.46) becomes



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Fig. 2.4 A nonideal transformer with small leakage.

$$A(\varepsilon) = \frac{1}{\varepsilon} \begin{bmatrix} -\frac{R_1}{L_1} & \frac{\sqrt{1-\varepsilon} R_2}{\sqrt{L_1 L_2}} \\ \frac{\sqrt{1-\varepsilon} R_1}{\sqrt{L_1 L_2}} & -\frac{R_2}{L_2} \end{bmatrix} \quad (2.48)$$

or, substituting $\sqrt{1-\varepsilon} = 1 - \frac{\varepsilon}{2} + O(\varepsilon^2)$

$$A(\varepsilon) = \frac{1}{\varepsilon} (A_0 + \varepsilon A_1(\varepsilon)) \quad (2.49)$$

where

$$A_0 = \begin{bmatrix} -\frac{R_1}{L_1} & \frac{R_2}{\sqrt{L_1 L_2}} \\ \frac{R_1}{\sqrt{L_1 L_2}} & -\frac{R_2}{L_2} \end{bmatrix} \quad (2.50)$$

The left null space of A_0

$$V = [L_1 \quad \sqrt{L_1 L_2}] \quad (2.51)$$

and the row space of A_0

$$W = [R_1 \quad -R_2 \sqrt{\frac{L_1}{L_2}}] \quad (2.52)$$

define, according to (2.45), the slow variable

$$y = L_1 x_1 + \sqrt{L_1 L_2} x_2 \quad (2.53)$$

and the fast variable

$$z = R_1 x_1 - R_2 \sqrt{\frac{L_1}{L_2}} x_2 \quad (2.54)$$

In the new coordinates y, z the state equations become

$$\begin{aligned}\frac{dy}{dt} &= - [1/(T_1+T_2)]y + [(T_1-T_2)/2(T_1+T_2)]z - \frac{v_1}{2} \\ \epsilon \frac{dz}{dt} &= \epsilon \left[\left(\frac{1}{T_2} - \frac{1}{T_1} \right) / 2 (T_1+T_2) \right] y - \left[\frac{1}{T_1} + \frac{1}{T_2} - \frac{\epsilon}{T_1+T_2} \right] z \\ &\quad - \left(\frac{1}{T_1} + \frac{1}{T_2} \right) v_1\end{aligned}\tag{2.55}$$

where $T_1 = L_1/R_1$, $T_2 = L_2/R_2$ are the time constants of the primary and secondary R-L circuits. The slow model is obtained by setting $\epsilon=0$ in the second equation of (2.55) and substituting the quasi-steady-state

$$\bar{z} = - v_1\tag{2.56}$$

into the first equation giving

$$\frac{d\bar{y}}{dt} = \left[- \frac{1}{T_1+T_2} \right] \bar{y} - \left[\frac{T_1}{T_1+T_2} \right] v_1.\tag{2.57}$$

The fast model is obtained by writing the second equation of (2.55) in the fast time-scale and setting $\epsilon=0$

$$\frac{d\tilde{z}}{d\tau} = - \left(\frac{1}{T_1} + \frac{1}{T_2} \right) \tilde{z} - \left(\frac{1}{T_1} + \frac{1}{T_2} \right) v_1.\tag{2.58}$$

It is interesting to note the physical interpretation of the new variables y, z and of (2.54). By writing

$$y = L_1 x_1 + \sqrt{L_1 L_2} x_2 = L_1 x_1 + M x_2 + 0(\epsilon) = \phi_{11} + \phi_{12} + 0(\epsilon)\tag{2.59}$$

we see that y is, to $O(\epsilon)$, the total flux linkage $\phi_{11} + \phi_{12}$ in coil 1. Aggregate physical quantities such as total flux linkage, total charge, total momentum etc., are often slow variables.

Noticing that $R_2 x_2$ is the voltage v_2 across the secondary winding of the transformer and using (2.54), (2.56) we obtain as $\epsilon \rightarrow 0$

$$v_2 = - (v_1 + R_1 x_1) \sqrt{\frac{L_2}{L_1}} = - (v_1 + R_1 x_1) \frac{N_2}{N_1} \quad (2.60)$$

where N_1, N_2 are the number of turns in coils 1 and 2. When the voltage drop $R_1 x_1$ is small compared to v_1 , which is usually the case, (2.60) reduces to the corresponding relation $v_2 = - \frac{N_2}{N_1} v_1$ for a leakage-free transformer.

Writing $M = \sqrt{L_1 L_2}$ $\sqrt{1-\epsilon} \simeq \sqrt{L_1 L_2} (1 - \frac{\epsilon}{2})$ we see that, for this example, Assumption 2.3 is satisfied.

2.6 Structured Singularly Perturbed Forms

Section 3 dealt with the nonexplicit singularly perturbed form (2.11), in which all the states are, generally, mixed. However, it is well known that a more structured system matrix implies that state x_1 is predominantly slow. In this section, the methodology developed in Section 3 is used to study two other structured forms, the fast separated form and the weak connection form.

A system is said to be in the fast separated singularly perturbed form if

$$\epsilon \begin{bmatrix} \frac{dx_1}{dt} \\ \frac{dx_2}{dt} \end{bmatrix} = \begin{bmatrix} \frac{dx_1}{d\tau} \\ \frac{dx_2}{d\tau} \end{bmatrix} = \begin{bmatrix} \epsilon A_{11} & A_{12} \\ \epsilon A_{21} & A_{22} \end{bmatrix} \quad (2.61)$$

where $x_1 \in \mathbb{R}^{n_1}$, $x_2 \in \mathbb{R}^{n_2}$, A_{11} , A_{12} , A_{21} , A_{22} are matrices of appropriate dimensions and A_{22} is a nonsingular matrix. Writing

$$\begin{bmatrix} \epsilon A_{11} & A_{12} \\ \epsilon A_{21} & A_{22} \end{bmatrix} = \begin{bmatrix} 0 & A_{12} \\ 0 & A_{22} \end{bmatrix} + \epsilon \begin{bmatrix} A_{11} & 0 \\ A_{21} & 0 \end{bmatrix} \quad (2.62)$$

we obtain

$$A_0 = \begin{bmatrix} 0 & A_{12} \\ 0 & A_{22} \end{bmatrix} \quad (2.63)$$

which satisfies Assumption 1. Since A_{22} is nonsingular the range and row spaces have dimensions n_2 whereas the left and right null spaces have dimension n_1 . It can be verified that

$$V = [I_{n_1} \quad -A_{12}A_{22}^{-1}] \quad (2.64)$$

$$W = [0 \quad I_{n_2}] \quad (2.65)$$

span the left null and row spaces of A_0 , respectively, and that

$$P = \begin{bmatrix} I_{n_1} \\ 0 \end{bmatrix}, \quad Q = \begin{bmatrix} A_{12} & A_{22}^{-1} \\ & I_{n_2} \end{bmatrix} \quad (2.66)$$

span the null and range spaces of A_0 , respectively. Moreover they satisfy (2.13).

Corollary 2.5 If A_{22} is nonsingular, then (i) the change of coordinates

$$y = x_1 - A_{12}A_{22}^{-1}x_2, \quad z = x_2 \quad (2.67)$$

transforms (2.61) into the explicit model (1.1) with $\bar{z}(t) = 0$. (ii) The slow reduced model of (2.61) is

$$\frac{d\bar{y}}{dt} = (A_{11} - A_{12}A_{22}^{-1}A_{21})\bar{y} \quad (2.68)$$

and the fast reduced model is

$$\frac{d\tilde{z}}{d\tau} = A_{22}\tilde{z} \quad (2.69)$$

(iii) The state x_2 of (2.61) is predominantly fast whereas x_1 is mixed.

Proof: (i) follows directly from Theorem 2.2, using (2.64), (2.65). (ii) is obtained by bearing in mind that $\bar{z}(t) = 0$. (iii) follows by inverting (2.67)

$$x_1 = y + A_{12}A_{22}^{-1}z, \quad x_2 = z \quad (2.70)$$

Note that state x_2 ($=z$) can be used in the fast reduced model (2.69) justifying the name "fast separated."

We now turn to the weak connection form which arises naturally in dynamic networks made of weakly connected "areas" [24-29]. A system is said to be in the weak connection form if^(*)

$$\epsilon \begin{bmatrix} \frac{dx_1}{dt} \\ \frac{dx_2}{dt} \end{bmatrix} = \begin{bmatrix} \frac{dx_1}{d\tau} \\ \frac{dx_2}{d\tau} \end{bmatrix} = \begin{bmatrix} A_{11} + \epsilon \hat{A}_{11} & \epsilon A_{12} \\ \epsilon A_{21} & A_{22} + \epsilon \hat{A}_{22} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} \quad (2.71)$$

where x_1 and x_2 are n_1 - and n_2 - vectors, A_{11} , A_{11} , A_{12} , A_{21} , A_{22} , A_{22} are matrices of appropriate dimensions and A_{11} , A_{22} are singular matrices with a complete set of eigenvectors corresponding to the zero eigenvalues, that is, satisfying

$$\mathcal{R}(A_{11}) \oplus \mathcal{N}(A_{11}) = \mathbb{R}^{n_1}, \quad \mathcal{R}(A_{22}) \oplus \mathcal{N}(A_{22}) = \mathbb{R}^{n_2} \quad (2.72)$$

with

$$\dim \mathcal{R}(A_{11}) = \rho_1 \geq 1, \dim \mathcal{N}(A_{11}) = \nu_1 \geq 1, \rho_1 + \nu_1 = n_1 \quad (2.73)$$

$$\dim \mathcal{R}(A_{22}) = \rho_2 \geq 1, \dim \mathcal{N}(A_{22}) = \nu_2 \geq 1, \rho_2 + \nu_2 = n_2. \quad (2.74)$$

Writing

$$\begin{bmatrix} A_{11} + \epsilon \hat{A}_{11} & \epsilon A_{12} \\ \epsilon A_{21} & A_{22} + \epsilon \hat{A}_{22} \end{bmatrix} = \begin{bmatrix} A_{11} & 0 \\ 0 & A_{22} \end{bmatrix} + \epsilon \begin{bmatrix} \hat{A}_{11} & A_{12} \\ A_{21} & \hat{A}_{22} \end{bmatrix} \quad (2.75)$$

(*) For convenience we deal with only two weakly connected "areas" but the ideas are directly applicable to any number of areas.

we obtain

$$A_0 = \begin{bmatrix} A_{11} & 0 \\ 0 & A_{22} \end{bmatrix} \quad (2.76)$$

which, because of (2.72), satisfies Assumption 2.1. Let V_i, W_i span the left null and row spaces of A_{ii} and P_i, Q_i span the null and range spaces, respectively, $i=1,2$, satisfying

$$V_i P_i = I_{V_i}, \quad W_i Q_i = I_{P_i}, \quad i=1,2. \quad (2.77)$$

Corollary 2.6 Under assumption (2.72), (i) the change of coordinates

$$\begin{bmatrix} y_1 \\ y_2 \end{bmatrix} = \begin{bmatrix} V_1 x_1 \\ V_2 x_2 \end{bmatrix}, \quad \begin{bmatrix} z_1 \\ z_2 \end{bmatrix} = \begin{bmatrix} W_1 x_1 \\ W_2 x_2 \end{bmatrix} \quad (2.78)$$

transforms (2.71) into the explicit model (1.1) with $\bar{z}(t) = 0$. (ii) The slow reduced model is

$$\frac{d}{dt} \begin{bmatrix} y_1 \\ y_2 \end{bmatrix} = \begin{bmatrix} V_1 \hat{A}_{11} P_1 & V_1 A_{12} P_2 \\ V_2 A_{21} P_1 & V_2 \hat{A}_{22} P_2 \end{bmatrix} \begin{bmatrix} y_1 \\ y_2 \end{bmatrix} \quad (2.79)$$

and the fast reduced model is

$$\frac{d}{d\tau} \begin{bmatrix} z_1 \\ z_2 \end{bmatrix} = \begin{bmatrix} W_1 A_{11} Q_1 + \epsilon W_1 \hat{A}_{11} Q_1 & \epsilon W_1 A_{12} Q_2 \\ \epsilon W_2 A_{21} Q_1 & W_2 A_{22} Q_2 + \epsilon W_2 \hat{A}_{22} Q_2 \end{bmatrix} \begin{bmatrix} z_1 \\ z_2 \end{bmatrix}. \quad (2.80)$$

Proof: (i) Apply Theorem 2.2 noting that

$$V = \begin{bmatrix} v_1 & 0 \\ 0 & v_2 \end{bmatrix}, \quad W = \begin{bmatrix} w_1 & 0 \\ 0 & w_2 \end{bmatrix} \quad (2.81)$$

span the left null and row spaces of A_0 , respectively. (ii) is obtained by keeping in mind that $\bar{z}(t) = 0$.

There are few interesting points to be noted. From (2.78) a slow vector y_1 and a fast vector z_1 are defined in terms of area x_1 only; similarly for x_2 . From (2.80) the fast variables z_1, z_2 are only weakly connected to each other and since $W_i A_{ii} Q_i$, $i=1,2$ are nonsingular these connections can be neglected for an $O(\epsilon)$ approximation. The fundamental difference between the original (2.71) and the transformed system (2.80) is that (2.80) no longer has a continuum of equilibrium points. Hence, each area defines a local fast model z_i connected with $O(\epsilon)$ connections to other local models, whereas contributions y_i from each area form a "slow core" describing the system-wide dynamics of (2.71). It is shown in Chapters 3 and 4 that this decomposition carries over to nonlinear weakly connected systems.

CHAPTER 3

NONEXPLICIT SINGULAR PERTURBATIONS IN NONLINEAR SYSTEMS

3.1 Introduction

An asymptotic procedure for time-scale separation in nonlinear models is of paramount importance since a nonlinear analog of the algebraic transformation in [37,38,41] is not available. In this chapter we give such a procedure and demonstrate its application to classes of nonlinear systems. The coordinate free characterization of singular perturbations [16] is extended to nonlinear systems of the form $\epsilon \dot{x} = h(x, \epsilon)$ and it is shown that equilibrium and conservation properties lead to a definition of new coordinates in which time scales are explicit (Section 3.2). In Section 3.3 we study a class of nonlinear high gain feedback control systems in which the controls enter linearly through a constant matrix but the open loop system, the output map and the feedback law may all be nonlinear. It is shown that these systems can be studied through singular perturbation techniques after they are transformed to the explicit form using the method of Section 3.2. The last section, 3.4, is devoted to interconnected systems whose isolated subsystems possess equilibrium and conservation properties. It is shown that in such systems weak connections give rise to two-time-scale behavior. Separation of the time scales defines a slow "core" which describes the system-wide behavior and a set of fast "residues" describing the local behavior of each subsystem. The decomposition, known for specific classes of linear models such as Markov chains [42-43] linearized models of power systems [24-28,44], electrical networks [44,45] and economic

systems [46] where it appeared for the first time, is established for a wide class of nonlinear systems with common features (i) equilibrium and conservation properties and (ii) weak connections between subsystems. Using this decomposition we give decentralized stability criteria for this class of systems, analogous to those in [47-49].

3.2 Conservation and Equilibrium Properties in Nonlinear Systems

The need for coordinate free characterization of time scales in nonlinear systems is more pressing than in linear systems. Wide separation of eigenvalues provides some characterization in linear systems but the notion of modes is nonexistent in nonlinear systems. It will be shown in this section that the conservation and equilibrium properties introduced in Chapter 2 for linear systems can naturally be extended to nonlinear systems and that they lead to a new set of variables in which the time scales are explicit.

To motivate the discussion we re-examine the explicit model (1.1) from a different point of view. Writing (1.1) in the fast time scale

$$\begin{aligned}\frac{dy}{d\tau} &= \epsilon f(y, z, \epsilon) \\ \frac{dz}{d\tau} &= g(y, z, \epsilon)\end{aligned}\tag{3.1}$$

and setting $\epsilon=0$ we obtain the auxiliary system

$$\begin{aligned}\frac{dy}{d\tau} &= 0 \\ \frac{dz}{d\tau} &= g(y, z, 0)\end{aligned}\tag{3.2}$$

which has the following two important properties.

Conservation Property A function of the state

$$\sigma(y, z) = y \quad (3.3)$$

remains at its initial value $\sigma(y(0), z(0)) = y(0)$, that is, it is conserved during the motion of (3.2).

Equilibrium Property System (3.2) possesses a set of nonisolated (continuum) equilibrium points defined by

$$g(y, z, 0) = 0. \quad (3.4)$$

The equilibria defined by (3.4) are the "quasi-steady-states" to which the fast transients of (3.1) converge as explained in Chapter 1.

A generalized version of (3.1) is a system in the form

$$\epsilon \frac{dx}{dt} = \frac{dx}{d\tau} = h(x, \epsilon) \quad (3.5)$$

which in τ -scale at $\epsilon=0$

$$\frac{dx}{d\tau} = h(x, 0) \quad (3.6)$$

has equilibrium and conservation properties analogous to the properties of (3.1). System (3.5) is studied in a domain $D \subset \mathbb{R}^n \times [0, \epsilon_0]$ in which function h is assumed to be continuously differentiable with respect to x and ϵ .

Assumption 3.1 System (3.5) satisfies the following conditions for existence of manifolds^(*) S and F .

Equilibrium Manifold S The set

$$S = \{x | h(x, 0) = 0, x \in D\} \quad (3.7)$$

defines a ν -dimensional differentiable manifold, $\nu \geq 1$. Hence, there exists continuously differentiable function $\varphi: R^n \rightarrow R^p$, $p = n - \nu$, $\text{rank } \frac{\partial \varphi}{\partial x} = p$, $\forall x \in D$ such that

$$\varphi(x) = 0 \Leftrightarrow h(x, 0) = 0 \quad (3.8)$$

that is, in the domain of interest D , every equilibrium of (3.6) satisfies $\varphi(x) = 0$ and every x satisfying $\varphi(x) = 0$ is an equilibrium of (3.6).

Dynamic Manifold F_{x_0} There exists continuously differentiable function $\sigma: R^n \rightarrow R^\nu$ such that for each $x(0) = x_0$ the ρ -dimensional ($\rho = n - \nu$) manifold

$$F_{x_0} = \{x | \sigma(x) - \sigma(x_0) = 0, \text{ rank } \frac{\partial \sigma}{\partial x} = \nu\} \quad (3.9)$$

is an invariant manifold of (3.6) that is a trajectory originating in F_{x_0} remains in F_{x_0}

$$\sigma(x(\tau)) - \sigma(x_0) = 0, \quad \forall \tau \geq 0. \quad (3.10)$$

^(*) Manifolds are generalizations to R^n of objects such as curves and surfaces in R^3 . More precisely, let $\tilde{\varphi}: R^n \rightarrow R^m$ be a continuously differentiable function from R^n into R^m . Then if the set $M = \{x | \tilde{\varphi}(x) = 0 \text{ and } \frac{\partial}{\partial x} \tilde{\varphi}(x) \text{ has rank } m\}$ is nonempty, it is an r -dimensional manifold, $r = n - m$ [50, 51].

Moreover, for all $x_0 \in D$, manifolds S and F_{x_0} are not tangent to each other, that is, for all x in the intersection of S and F_{x_0}

$$\text{rank} \begin{bmatrix} \frac{\partial \varphi}{\partial x} \\ \frac{\partial \sigma}{\partial x} \end{bmatrix} = \eta \quad (3.11)$$

Theorem 3.2 Under assumption 3.1, the change of coordinates

$$y = \sigma(x), \quad z = \varphi(x) \quad (3.12)$$

transforms (3.5) into the separated explicit model (1.1) with $\left. \frac{\partial g}{\partial z} \right|_{\varepsilon=0}$ nonsingular and $\bar{z}(t) = 0$.

Proof: Differentiating (3.10) and using (3.6) we have

$$\frac{\partial \sigma}{\partial x} h(x, 0) = 0 \quad (3.13)$$

Differentiating $y = \sigma(x)$ and using the mean value theorem in ε for each component of h

$$\frac{dy}{dt} = \frac{1}{\varepsilon} \frac{\partial \sigma}{\partial x} h(x, \varepsilon) = \frac{\partial \sigma}{\partial x} \frac{\partial h}{\partial \varepsilon} \quad (3.14)$$

we see that y is the slow variable of (1.1). Using the inverse transformation $x = \gamma(y, z)$ of (3.12) which exists because of (3.11), and differentiating $z = \varphi(x)$ we obtain

$$\varepsilon \frac{dz}{dt} = \frac{\partial \varphi}{\partial x} h(x, \varepsilon) = \frac{\partial \varphi}{\partial x} h(\gamma(y, z), \varepsilon) \triangleq g(y, z, \varepsilon) \quad (3.15)$$

We show that $(\partial g / \partial z) \big|_{\epsilon=0}$ is nonsingular by contradiction. Assuming that it is singular the equilibrium manifold of (3.14)-(3.15) has dimension greater than ν which is contradiction because (3.12) is a nonsingular transformation. Finally from $x \in S \Leftrightarrow x = \gamma(y, 0)$ it follows that $h(\gamma(y, 0), 0) = 0$ and

$$g(y, 0, 0) = 0 \quad (3.16)$$

implying that $\bar{z}(t) = 0$.

The intuitive idea behind this theorem is illustrated by Fig. 3.1. If the equilibrium manifold S is attractive, the trajectories of (3.6) which are confined to some F due to (3.10), converge to S and when $\tau \rightarrow \infty$ they terminate at the intersection of F and S . Instead the trajectories of (3.5) rapidly approach S staying in a boundary layer close to F and then slowly continue their motion remaining close to S . Since the trajectories are initially close to F the quantity $\sigma(x)$ stays almost constant during this interval; thus it qualifies as a predominantly slow variable. On the other hand, the quantity $\varphi(x)$, which is large away from S where the trajectory starts, rapidly diminishes when the trajectory approaches S ; thus it qualifies as a predominantly fast variable.

As an illustration we consider (3.17)

$$\begin{aligned} \frac{dx_1}{d\tau} &= -\varphi_1(x) + (x_1 + x_3) \varphi_2(x) - \epsilon x_1 \\ \frac{dx_2}{d\tau} &= -2x_2 \varphi_2(x) - \epsilon x_2^3 \\ \frac{dx_3}{d\tau} &= \varphi_1(x) + (x_1 + x_3) \varphi_2(x) - \epsilon x_3 \end{aligned} \quad (3.17)$$

over $R^3 \supset D = \{(x_1, x_2, x_3) \mid x_1 > 1, x_2 > 0.5, x_3 > 0.5\}$ where $\varphi_1(x), \varphi_2(x)$ are continuously differentiable functions defined over D . Setting $\varepsilon=0$ in (3.17) we obtain

$$\begin{aligned}\frac{dx_1}{d\tau} &= -\varphi_1(x) + (x_1+x_3)\varphi_2(x) \\ \frac{dx_2}{d\tau} &= -2x_2\varphi_2(x) \\ \frac{dx_3}{d\tau} &= \varphi_1(x) + (x_1+x_3)\varphi_2(x)\end{aligned}\tag{3.18}$$

for which

$$\varphi_1(x)=0, \quad \varphi_2(x)=0\tag{3.19}$$

define the equilibrium manifold S . It is easily verified that the dynamic manifolds are defined by $\sigma(x) = \sigma(x(0))$ where

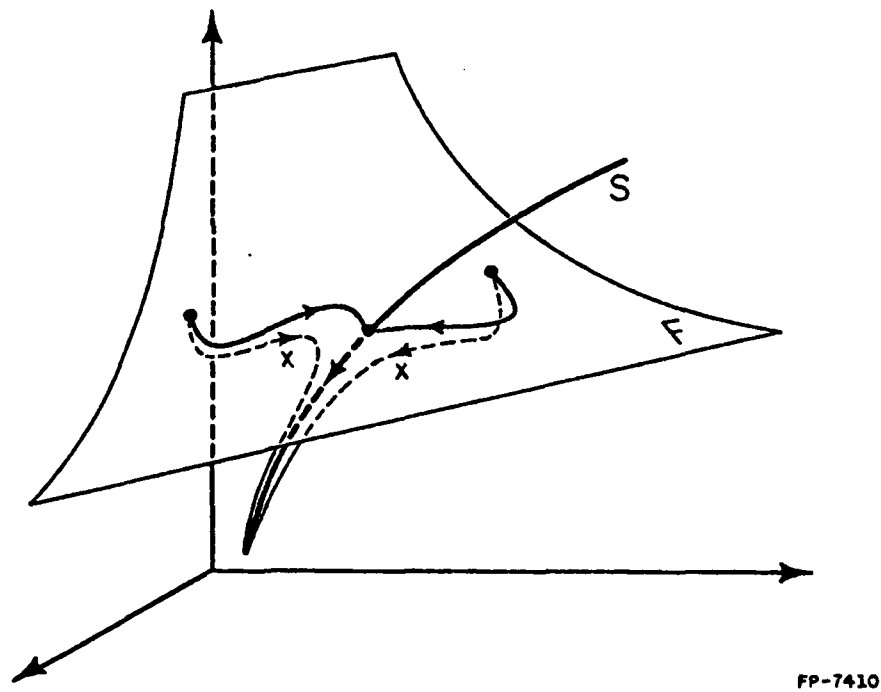
$$\sigma(x) = (x_1+x_3)x_2.\tag{3.20}$$

The equilibrium manifold and a dynamic manifold of this system are shown in Fig. 3.1 where functions φ_1, φ_2 were chosen as

$$\varphi_1(x) = x_1 - x_3, \quad \varphi_2(x) = x_2^2 - x_3 - x_1 + 1.\tag{3.21}$$

3.3 High Gain Feedback and Disturbance Rejection

Use of high gain in feedback loops has been known to reduce the effects of disturbances, parameter variations and distortions [19,52]. Early investigations using root locus techniques [53] have shown that



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Fig. 3.1 Equilibrium (S) and dynamic (F) manifolds of (3.18).

under high gain some poles of the closed loop system tend to infinity, a characteristic of singularly perturbed systems. Similar behavior is exhibited by multivariable systems [54]. An extensive study of high gain in Linear Time Invariant (LTI) systems was undertaken in [17,55] where it was pointed out that every high gain system is a singularly perturbed one and vice versa. In this section we show that the relation between high gain feedback and singular perturbations extends beyond the class of LTI systems.

We consider the system

$$\begin{aligned}\dot{x} &= f(x) + Bu \\ y &= g(x)\end{aligned}\tag{3.22}$$

under the output feedback

$$u = \frac{1}{\varepsilon} k(y)\tag{3.23}$$

and we study the behavior of the closed loop system when the gain $\frac{1}{\varepsilon} \rightarrow \infty$.

The state x is a n -vector and the input and output vectors both have dimension $m < n$. Functions f, g are defined in a domain $D_x \subset \mathbb{R}^n$ and function k is defined in a domain $D_y \subset \mathbb{R}^m$. All functions are assumed to be differentiable a sufficient number of times. Moreover we make the following basic assumption.

Assumption 3.3 (a) Matrix B is full rank. Hence, there exists a $v \times n$ ($v=n-m$) matrix V , rank $V=v$, such that $VB=0$.

(b) There exists a unique set point $y^* \in D_y$ such that

$$k(y^*) = 0\tag{3.24}$$

(c) Matrix $\begin{bmatrix} V \\ g_x \end{bmatrix}$, where g_x is the partial derivative of g , is nonsingular $\forall x \in D_x$ satisfying $g(x) - y^* = 0$.

Part b of the assumption could be relaxed to allow isolated roots of (3.24). However, nonisolated roots which imply dead zones are excluded. When the output is linear in x , $y=Cx$, Assumption 3.3(c) requires $\begin{bmatrix} V \\ C \end{bmatrix}$ to be nonsingular which is equivalent to the assumption of [17,55] that CB is nonsingular. Indeed if Assumption 3.3(c) holds $\begin{bmatrix} V \\ C \end{bmatrix} B = \begin{bmatrix} 0 \\ CB \end{bmatrix}$ must be full rank and $(CB)^{-1}$ exists; conversely if $(CB)^{-1}$ exists the row space of C and the left null space of B are disjoint and $\begin{bmatrix} V \\ C \end{bmatrix}$ is nonsingular.

Theorem 3.4 Under Assumption 3.3 the transformation

$$\begin{aligned} y_s &= Vx & (*) \\ z &= g(x) - y^* \end{aligned} \quad (3.25)$$

transforms the high gain system (3.22)-(3.23) into an explicit singularly perturbed form with $\bar{z}=0$.

Proof: Substituting (3.23) into (3.22) and rescaling the time, $\tau=t/\epsilon$ we obtain

$$\frac{dx}{d\tau} = \epsilon f(x) + Bk(g(x)) \quad (3.26)$$

whose auxiliary system is

$$\frac{dx}{d\tau} = Bk(g(x)) \quad (3.27)$$

* We temporarily change notation letting v_s instead of y denote the slow variable. The latter has been reserved for its more traditional use as an output variable.

Since by Assumption 3.3(a)

$$V \frac{dx}{d\tau} = VBk(g(x)) = 0 \quad (3.28)$$

the relation

$$\sigma(x) = Vx, \quad \sigma(x) = \sigma(x(0)) \quad (3.29)$$

defines the family of m-dimensional dynamic manifolds. By Assumption 3.3(b) all $x \in D_x$ satisfying

$$g(x) = y^* \quad (3.30)$$

are equilibria of (3.27). Assumption 3.3(c) implies that (3.30) defines a (n-m)-dimensional equilibrium manifold which is transversal to the dynamic manifolds. The transformation is simply an application of Theorem 3.2.

In the new coordinates (3.22)-(3.23) becomes

$$\begin{aligned} \frac{dy_s}{dt} &= Vf(\gamma(y_s, z)) \\ \epsilon \frac{dz}{dt} &= \epsilon g_x f(\gamma(y_s, z)) + g_x Bk(z + y^*) \end{aligned} \quad (3.31)$$

where $x = \gamma(y_s, z)$ is the inverse transformation to (3.25) which exists because of Assumption 3.3(c).

Corollary 3.5 If (i) Assumption 3.3 is satisfied and (ii) the boundary layer system

$$\left. \frac{dz}{d\tau} = g_x \right|_{x=\gamma(y_s, z)} Bk(\hat{z} + y^*) \quad (3.32)$$

is asymptotically stable, the response of (3.22)-(3.23) is approximated, over a bounded interval $[0, T]$, by

$$\begin{aligned} x(t) &= \gamma(\bar{y}_s(t), \tilde{z}(\frac{t}{\epsilon}) + 0(\epsilon) \\ y(t) &= \tilde{z}(\frac{t}{\epsilon}) + y^* + 0(\epsilon) \end{aligned} \quad (3.33)$$

where $\bar{y}_s(t)$ satisfies the reduced system

$$\frac{dy_s}{dt} = Vf(\gamma(\bar{y}_s, 0)) \quad (3.34)$$

Proof: Follows immediately from Theorem 3.4 and standard singular perturbation results.

From (3.33) the output differs from the set point y^* by the predominantly fast variable $\tilde{z} + 0(\epsilon)$. Hence, any disturbance that can be modelled as initial condition will appear in the output only over a short initial interval. Note, however, that with the assumptions in Corollary 3.5 the approximation (3.33) is valid only over a bounded time interval $[0, T]$. Under stronger conditions, which essentially amount to stability requirements on the slow system (3.34), the approximation is valid over unbounded intervals [36] and disturbance rejection is indeed achieved.

Note also that the dynamic manifold defined in (3.29) is linear because we assumed that the input enters linearly through a constant matrix. If in addition the output is linear in x , the equilibrium manifold is linear too and (3.25) is a linear transformation. In this case the boundary layer system (3.32) depends on \tilde{z} only and application of

Theorem 3.4 and Corollary 3.5 is greatly facilitated since transformation (3.25) can be inverted explicitly and stability of (3.32) can be checked more easily. The following example illustrates the discussion above.

Example 3.6 The output of

$$\begin{aligned} \dot{x}_1 &= f_1(x) + u \\ \dot{x}_2 &= f_2(x) + u \end{aligned} \quad x = \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} \quad (3.35)$$

$$y = x_1^2 + x_2 \quad (3.36)$$

is fed back through the high gain law

$$u = \frac{1}{\epsilon} k(y) = \frac{1}{\epsilon} (-y - y^3) \quad (3.37)$$

This system satisfies Assumption 3.3(a), 3.3(b) and

$$V = [1 \ -1], \quad y^* = 0 \quad (3.38)$$

It also satisfies Assumption 3.3(c) if we restrict the region of validity to $x_1 \geq 0.6$. Transformation (3.25) becomes

$$\begin{aligned} y_s &= x_1 - x_2 \\ z &= x_1^2 + x_2 \end{aligned} \quad (3.39)$$

with inverse

$$\begin{aligned} x_1 &= \frac{-1 \pm \sqrt{1+4(y_s+z)}}{2} \\ x_2 &= \frac{-1 \pm \sqrt{1+4(y_s+z)}}{2} - y_s \end{aligned} \quad (3.40)$$

and the transformed system is

$$\begin{aligned} \frac{dy_s}{dt} &= [f_1(x) - f_2(x)] \Big|_{x=\gamma(y_s, z)} \\ \epsilon \frac{dz}{dt} &= \epsilon [2x_1 f_1(x) + f_2(x)] \Big|_{x=\gamma(y_s, z)} - [\pm \sqrt{1+4(y_s+z)}] (z+z^3) . \end{aligned} \quad (3.41)$$

Since the boundary layer system

$$\frac{dz}{d\tau} = - \sqrt{1+4(y_s+z)} (z+z^3) \quad (3.42)$$

is asymptotically stable in the region of validity, approximation (3.33) holds for large enough gain.

Transformation (3.25) can also be used to analyze disturbance rejection when the disturbance is modelled as an input. Consider

$$\begin{aligned} \dot{x} &= f(x, w(t)) + Bu \\ y &= g(x) \end{aligned} \quad (3.43)$$

under the output feedback

$$u = \frac{1}{\epsilon} k(y) \quad (3.44)$$

whose difference from (3.22)-(3.23) is that the disturbance $w(t)$ appears as input to the system. Substituting (3.44) into (3.43) we obtain

$$\epsilon \frac{dx}{dt} = \epsilon f(x, w(t)) + Bk(g(x)) \quad (3.45)$$

which, under Assumption 3.3, can be transformed to

$$\frac{dy_s}{dt} = Vf(\gamma(y_s, z), w(t)) \triangleq \hat{f}(t, y_s, z) \quad (3.46)$$

$$\epsilon \frac{dz}{dt} = \epsilon g_x f(\gamma(y_s, z), w(t)) + g_x Bk(z+y^*) \triangleq \hat{g}(t, y_s, z, \epsilon)$$

using (3.25). In the following corollary we make use of a theorem in [36] which, for convenience is reproduced in Appendix I.

Corollary 3.7 Assume that $w(t)$ is such that \hat{f}, \hat{g} , satisfy all the conditions of the theorem in [36]. Then under Assumption 3.3 and for high enough gain $\frac{1}{\epsilon}$ the output of (3.43) remains $O(\epsilon)$ close to the set point y^* for all $t \in [0, \infty)$.

Proof: Since the boundary layer system of (3.46)

$$\frac{d\tilde{z}}{d\tau} = g_x Bk(\tilde{z}+y^*) \quad (3.47)$$

has $\tilde{z}=0$ as its unique equilibrium and $y(0)=y^*$ implies $z(0)=y(0)-y^*=0$, the fast part of z is $\tilde{z}(\tau)=0 \forall t \in [0, \infty)$. Furthermore, $\bar{z}(t)=0$. Hence,

$$y(t) = z(t) + y^* = y^* + O(\epsilon) \quad (3.48)$$

by the theorem in [36].

The practical use of the above corollary may be limited since it assumes that $w(t)$ is known so that the assumptions in [36] can be checked. A much more desirable result would be to establish (3.48) for a class of inputs $w(t)$. Such a result should draw upon the specific way in which the disturbance enters into the problem.

3.4 Interconnected Systems

An appealing approach to large scale system analysis and design is to view the system as a collection of dynamic subsystems interacting through static interconnections. The object then is to analyze the stability of [47-49] or design control laws for the system [56] in a decentralized fashion, that is, by testing the stability of the subsystems or designing feedback control using only the subsystem states or outputs. This approach is based on the premise that the connections between subsystems are "weak" compared to the internal connections; hence, qualitative properties and control design can be performed on the subsystem level.

In this section we show that when subsystems have equilibrium and conservation properties, weak connections give rise to two-time-scale behavior. Subsystems are weakly coupled* in the fast time scale but are strongly coupled in the slow one. We start by showing that such cases arise naturally in high gain decentralized output feedback. Consider the interconnected system

$$\begin{aligned}\dot{x}_i &= \hat{f}_i(x_i) + \hat{h}_i(x) + B_i u_i \\ y_i &= \hat{g}_i(x_i) \qquad i=1, \dots, l \\ x &= [x_1^T \dots x_l^T]^T\end{aligned}\tag{3.49}$$

with decentralized output feedback

*We use the word "connection" to mean physical, static interaction and the word "coupling" to imply dynamic interaction.

$$u_i = \frac{1}{\epsilon} \hat{k}_i(y_i) \quad i=1, \dots, l \quad (3.50)$$

where B_i , \hat{g}_i , \hat{k}_i satisfy Assumption 3.3, $i=1, \dots, l$. Substituting (3.50) into (3.49) and rescaling the time $\tau = \frac{t}{\epsilon}$ we obtain

$$\frac{dx_i}{d\tau} = B_i \hat{k}_i(\hat{g}_i(x_i)) + \epsilon(\hat{f}_i(x_i) + \hat{h}_i(x)) \quad i=1, \dots, l \quad (3.51)$$

whose subsystems

$$\frac{dx_i}{d\tau} = B_i \hat{k}_i(\hat{g}_i(x_i)) \quad i=1, \dots, l \quad (3.52)$$

have equilibrium and conservation properties as argued in section 3.3.

Generalizing (3.51) we consider weakly interconnected systems of the form

$$\frac{dx_i}{d\tau} = f_i(x_i, \epsilon) + \epsilon g_i(x, \epsilon) \quad i=1, \dots, l \quad (3.53)$$

where f_i is defined on a domain $D_i \times [0, \epsilon_i] \subset \mathbb{R}^{n_i} \times \mathbb{R}$, g_i is defined on $D \times [0, \epsilon] \subset \mathbb{R}^n \times \mathbb{R}$, $x = [x_1^T \dots x_l^T]^T$ and $n = \sum_{i=1}^l n_i$. Functions f_i and g_i are assumed to be sufficiently smooth. In (3.53) $f_i(x_i, \epsilon)$ represents the i th isolated subsystem [47] whereas, $\epsilon g_i(x, \epsilon)$ represents interconnections with other subsystems. Although the dependence of f_i on ϵ may seem superfluous, it is sometimes needed to assure the existence of equilibria of the isolated subsystems; such a case is discussed in the next chapter in relation to dynamic networks. Concerning the isolated subsystems we make the following assumption.

Assumption 3.8 Every isolated subsystem

$$\frac{dx_i}{d\tau} = f_i(x_i, 0) \quad , \quad i=1, \dots, l \quad (3.54)$$

has equilibrium and conservation properties. That is

(i) the set

$$S_i = \{x_i \mid f_i(x_i, 0) = 0\} \quad (3.55)$$

is a v_i -dimensional equilibrium manifold of (3.54), $0 \leq v_i \leq n_i$; hence, there exists smooth function $\varphi_i: R^{n_i} \rightarrow R^{\rho_i}$, $\rho_i = n_i - v_i$, such that

$$\varphi_i(x_i) = 0 \iff f_i(x_i, 0) = 0 \quad (3.56)$$

(ii) there exists function $\sigma_i: R^{n_i} \rightarrow R^{v_i}$ such that

$$F_i = \{x_i \mid \sigma_i(x_i) = \sigma_i(x_i(0))\} \quad (3.57)$$

is a family of invariant manifolds of (3.54) parametrized on $\sigma_i(x_i(0))$.

Moreover, S_i and F_i are nontangent, i.e.

$$\text{rank} \begin{bmatrix} \varphi_{ix} \\ \sigma_{ix} \end{bmatrix} = n, \quad \forall x_i \in D_i \quad (3.58)$$

where $\varphi_{ix}, \sigma_{ix}$ are the Jacobian matrices of φ_i, σ_i .

Corollary 3.9 Under Assumption 3.8 the interconnected system (3.53) is a two-time-scale system and the transformation

$$\begin{aligned} y_i &= \sigma_i(x_i) \\ z_i &= \varphi_i(x_i) \end{aligned} \quad i=1, \dots, l \quad (3.59)$$

transforms (3.53) into an explicit model with $v = \sum_{i=1}^l v_i$ predominantly slow variables y and $\rho = \sum_{i=1}^l \rho_i$ predominantly fast variables z for which $\bar{z}=0$.

Proof: Isolated subsystems (3.54) form the auxiliary system of (3.53) obtained by setting $\epsilon=0$.

Defining

$$\varphi(x) = \begin{bmatrix} \varphi_1(x_1) \\ \varphi_2(x_2) \\ \vdots \\ \varphi_l(x_l) \end{bmatrix}, \quad \sigma(x) = \begin{bmatrix} \sigma_1(x_1) \\ \sigma_2(x_2) \\ \vdots \\ \sigma_l(x_l) \end{bmatrix} \quad (3.60)$$

we obtain equilibrium manifold $\varphi(x)=0$, and dynamic manifolds $\sigma(x(\tau))=\sigma(x(0))$ of the interconnected system satisfying Assumption 3.1. The conclusion is then an application of Theorem 3.2.

Note that the dimension of manifold S_i of each subsystem is allowed to take extreme values 0 and n_i . When $v_i=0$ (3.54) has at the most isolated equilibria and its dynamic manifold is the whole space R^{n_i} ; when $v_i=n_i$ (3.54) is made of n_i integrators and its equilibrium manifold is R^{n_i} . Note also that transformation (3.59) is block diagonal in the sense that y_i and z_i are defined in terms of subsystem state x_i only.

Noting that

$$x_i = \gamma_i(y_i, z_i) \quad (3.61)$$

$$x = \gamma(y, z) = [\gamma_1(y_1, z_1) \ \dots \ \gamma_l(y_l, z_l)]$$

is the inverse transformation of (3.59) which exists due to (3.58) and rescaling the time $t=\epsilon\tau$ the transformed model is

$$\begin{aligned} \frac{dy_i}{dt} &= \sigma_{ix} \left[\frac{\partial f_i(\gamma_i(y_i, z_i), \epsilon)}{\partial \epsilon} + g_i(\gamma(y, z), \epsilon) \right] \triangleq F_i(y, z, \epsilon) \\ \epsilon \frac{dz_i}{dt} &= \varphi_{ix} [f_i(\gamma_i(y_i, z_i), \epsilon) + \epsilon g_i(\gamma(y, z), \epsilon)] \triangleq G_i(y_i, z_i, \epsilon) + \epsilon H_i(y, z, \epsilon) \end{aligned} \quad (3.62)$$

where $\sigma_{ix}, \varphi_{ix}$ are the partial derivatives of σ_i, φ_i with respect to x_i . According to Corollary 3.9 the quasi-steady-state is $\bar{z}=0$, and the slow model is

$$\frac{d\bar{y}_i}{dt} = F_i(\bar{y}, 0, 0) \quad i=1, \dots, l \quad (3.63)$$

Rescaling back to τ and setting $\epsilon=0$ in the second equation of (3.62) we obtain the fast model

$$\frac{d\tilde{z}_i}{d\tau} = G_i(y_i, \tilde{z}_i, 0) \quad i=1, \dots, l \quad (3.64)$$

where y_i appears as a parameter.

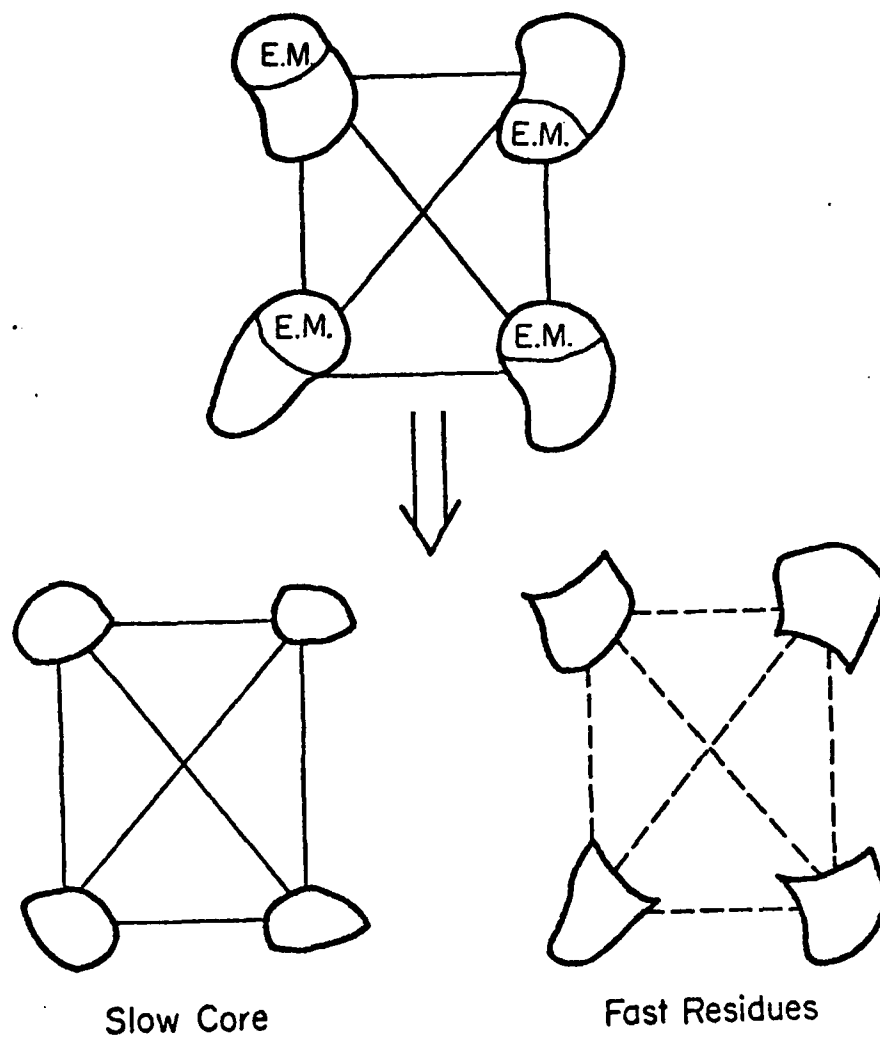
Note that F_i is a function of the whole \bar{y} vector whereas, G_i is a function of \tilde{z}_i only. Hence, separation of time scales has resulted in a decomposition in which parts from every subsystem are put together to form a slow core (y -variables) while the rest of each subsystem forms a fast residue (z_i -variables). The slow core describes the system-wide dynamics which due to the weak connections between subsystems become significant only in the long run. The fast residues describe the local dynamics which, due to the strong connections within subsystems, are significant in the short run. If further, the fast residues are asymptotically stable the z_i variables reach quickly their quasi-steady-state

equilibrium ($\bar{z}_i=0$); hence, they are weakly coupled with each other since interaction from subsystem to subsystem through the weak connections becomes noticeable only in the slow time scale. (See (3.64)). Figure 3.2 gives a pictorial view of the discussion. This decomposition is very reminiscent of Simon and Ando's reasoning in their classical 1961 paper [46]. We quote:

- (1) We can somehow classify all the variables in the economy into a small number of groups;
- (2) We can study the interactions within the groups as though the interaction among groups did not exist;
- (3) We can define indices representing groups and study the interaction among these indices without regard to the interactions within each group.

Step (1) corresponds in our case to identifying subsystems connected to each other through weak connections. Step (2) corresponds to our fast models (3.64) which are disconnected; we went one step further to remove the slow motion from each subsystem. Step (3) corresponds to the definition of slow variables y_i as "indices" representing subsystems and the study of the system-wide dynamics through the slow core (3.63). In the next chapter where the decomposition is specialized to dynamic networks, the slow "indices" take the meaning of aggregate physical variables.

There is an extensive literature devoted to stability analysis of interconnected systems [47-49 and references therein]. The general plan followed is (i) to regard the large scale system as an interconnection



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Fig. 3.2 The decomposition into slow core and fast residues.

of isolated subsystems, (ii) to characterize stability properties of isolated subsystems through Lyapunov techniques, (iii) to deduce stability properties of the overall system from stability of subsystems and the nature of interconnections. A basic assumption in [47,48] is that the subsystems have isolated equilibria and the results of [47,48] do not directly apply when the subsystems have nonisolated (i.e., a continuum of) equilibrium points such as in (3.53) with $f_i(x_i, 0)$ satisfying Assumption 3.8. We now show that stability criteria analogous to those in [47] can be derived based on recent results [20] on stability of singularly perturbed systems. For convenience, these results are reproduced in Appendix II.

We consider the interconnected system (3.53) when $f_i(x_i, 0)$ satisfies Assumption 3.8. Using transformation (3.59) we obtain the system in the explicit singularly perturbed form (3.62). We assume that $y_i \in D_{y_i} \subset R^{v_i}$, $z_i \in D_{z_i} \subset R^{p_i}$, $i=1, \dots, l$ and that $y=0, z=0$ is the unique equilibrium of (3.62). Moreover the following assumptions are made concerning the slow core, the fast residues and their interactions.

Assumption 3.10

(i) The slow core (3.63) has a Lyapunov function $V: R^v \rightarrow R_+$ such that for all $y \in D_y$

$$[\nabla_y V(y)]^T F(y, 0, 0) \leq -\alpha_1 \psi^2(y), \quad \alpha_1 > 0$$

where $\psi(y)$ is a scalar valued function of y with $\psi(0)=0$ and $\psi(y) \neq 0$, $y \neq 0$.

(ii) Every isolated fast residue (3.64) has a Lyapunov function

$W_i(y_i, z_i): R^{v_i} \times R^{p_i} \rightarrow R_+$ such that for all $y_i \in D_{y_i}, z_i \in D_{z_i}$

$$[\nabla_{z_i} W_i(y_i, z_i)]^T G_i(y_i, z_i, 0) \leq -\alpha_{2i} \phi_i^2(z_i) \quad \alpha_{2i} > 0.$$

(iii) There exists $\lambda_i > 0$, $i=1, \dots, l$ such that for all $y \in D_y$, $z \in D_z$ the following hold:

$$\begin{aligned} (a) \quad & [\nabla_{y_i} W_i(y_i, z_i)]^T F_i(y, z, \varepsilon) \leq c_i \sum_{i=1}^l \lambda_i \Phi_i^2(z_i) \\ (b) \quad & [\nabla_y V(y)]^T [F(y, z, \varepsilon) - F(y, 0, \varepsilon)] \leq \mathcal{B} \Psi(y) (\sum \lambda_i \Phi_i^2(z_i))^{1/2} \\ (c) \quad & [\nabla_{z_i} W_i(y_i, z_i)]^T [G_i(y_i, z_i, \varepsilon) - G_i(y_i, z_i, 0)] \leq \varepsilon k_{1i} \Phi_i^2(z_i) \\ (d) \quad & [\nabla_{z_i} W_i(y_i, z_i)]^T H_i(y, z, \varepsilon) \leq k_2 \sum_{i=1}^l \lambda_i \Phi_i^2(z_i). \end{aligned}$$

Constants $c_i, \mathcal{B}, k_{1i}, k_2$ are all assumed to be nonnegative numbers.

Theorem 3.11 If Assumption 3.10 is satisfied and ε is sufficiently small the equilibrium ($y=0, z=0$) of (3.62) is asymptotically stable.

Proof: Let

$$W(y, z) = \sum_{i=1}^l \lambda_i W_i(y_i, z_i) \quad (3.65)$$

be a tentative Lyapunov function for the boundary layer system, formed by the z_i -systems, $i=1, \dots, l$, in (3.64). Then,

$$\begin{aligned} [\nabla_z W(y, z)]^T G(y, z, 0) &= \sum_{i=1}^l \lambda_i \nabla_{z_i} W_i(y_i, z_i) G_i(y_i, z_i, 0) \\ &\leq - \sum_{i=1}^l \lambda_i a_{2i} \Phi_i^2(z_i) \leq - \alpha_2 \Phi^2(z) \end{aligned} \quad (3.66)$$

where

$$\Phi^2(z) = \sum_i \lambda_i \Phi_i^2(z_i) \quad , \quad \alpha_2 = \min_i \alpha_{2i} \quad (3.67)$$

Hence, $V(y), W(y, z)$ satisfy condition (I), (II) of [20]. Condition IIIa of [20] is also satisfied since

$$\begin{aligned}
[\nabla_y W(y, z)]^T F(y, z, \epsilon) &= \sum_i \lambda_i [\nabla_{y_i} W_i(y_i, z_i)]^T F_i(y, z, \epsilon) \\
&\leq \sum_i \lambda_i c_i \phi^2(z) = (\sum_i \lambda_i c_i) \phi^2(z)
\end{aligned} \tag{3.68}$$

whereas condition IIIb is identical to (iiib) in Assumption 3.10.

Finally, letting

$$\hat{G}_i(y, z, \epsilon) = G_i(y_i, z_i, \epsilon) + \epsilon H_i(y, z, \epsilon) \tag{3.69}$$

we have

$$\begin{aligned}
&[\nabla_z W(y, z)]^T [\hat{G}(y, z, \epsilon) - \hat{G}(y, z, 0)] = \\
&\sum_i [\nabla_{z_i} W_i(y_i, z_i)]^T [G_i(y_i, z_i, \epsilon) - G_i(y_i, z_i, 0) + \epsilon H_i(y, z, \epsilon)] \leq \\
&\epsilon \sum_i k_{1i} \phi_i^2(z_i) + \epsilon k_2 \phi^2(z) \leq \epsilon K_1 \phi^2(z)
\end{aligned} \tag{3.70}$$

where

$$K_1 = \max_i \frac{k_{1i}}{\lambda_i} + k_2 \tag{3.71}$$

and we see that condition IIIc of [20] is also satisfied. The conclusion follows directly from [20].

A similar procedure cannot be applied to the original model (3.53) because the isolated systems possess a continuum of equilibrium points. The basic difference between (3.53) and (3.62) is that in (3.62) all the slow dynamics giving rise to equilibrium manifolds have been relegated to the slow core; consequently, the fast residues no longer have a continuum of equilibrium points. The stability criteria become easy to apply when the transformed models are structured, such as in dynamic networks, the subject of the next chapter.

CHAPTER 4

REDUCED ORDER MODELING OF DYNAMIC NETWORKS

4.1 Introduction

Much of the equilibrium-conservation reasoning presented in Chapters 2 and 3 was inspired by the study of time scales in power systems and other weakly connected networks. In this class of systems, which is the subject of the present chapter, states and connections have physical meaning and separation of time scales is related to physical laws such as conservation of mass, charge, momentum, etc.

Section 4.2 discusses weakly connected networks with linear storage but nonlinear interconnection elements. The main result is that the transformation that brings the system into the explicit singularly perturbed form is linear. In the new coordinates a slow core describes the system-wide behavior while fast residues describe the local behavior of the network. The slow core turns out to be another "aggregate" network whose states and connections are related in an intuitively appealing way to the states and connections of the original. These results are then specialized to power systems and a five-machine example illustrates the reduction procedure (section 4.3); simulation results are also shown. In section 4.4 we clarify the relation between coherency and localizability [24-25] for LTI systems.

4.2 Time Scales in Nonlinear Dynamic Networks

The time-scale separation methodology developed in the previous chapter will now be applied to nonlinear dynamic networks, a class of large systems whose structure facilitates the derivation of reduced models with physical meaning. The dynamic networks we consider are systems comprised of storage elements, capable of storing some physical quantity and interconnection elements capable of transporting this quantity without delay. Examples of dynamic networks include power systems, where angular momentum stored in the generators is transported through transmission lines, R-C networks where charge in the capacitors is transported through resistors, mass-spring systems, etc. The dynamic networks considered first have storage elements with linear characteristics but their interconnections may be nonlinear. Extension to networks with nonlinear storage elements is indicated later in the section. The rates of flow in the interconnections are assumed to be continuously differentiable functions of the potential differences across the interconnections satisfying

$$f_{ik}(x_i - x_k) = - f_{ki}(x_k - x_i) \quad . \quad (4.1)$$

This assumption is equivalent to saying that there are neither sources nor sinks along the interconnection. The dynamics of these systems are then modeled by either the system of first order equations

$$\dot{x}_i = - \frac{1}{m_i} \left[\sum_{k \in k_i} f_{ik}(x_i - x_k) - I_i \right] \quad (4.2)$$

or the system of second order equations

$$\ddot{x}_i = - \frac{1}{m_i} \left[\sum_{k \in k_i} f_{ik}(x_i - x_k) - I_i \right] \quad (4.3)$$

where x_i, m_i the potential and inertia of the i th storage element, I_i the net injection at the i th element, $f_{ik}(\cdot)$ the characteristic of the interconnection between elements i and j and K_i the set of elements to which i is connected. In the remainder of this section we deal with dynamic networks in the form (4.2). In the next section the results will be applied to power systems whose equations are in the form (4.3).

A dynamic network is said to be weakly connected if some interconnections can be expressed as multiples of a small parameter ϵ . The model of a weakly connected network is then

$$\frac{dx_i}{d\tau} = -\frac{1}{m_i} \left[\sum_{k \in K_i} f_{ik}(x_i - x_k) + \epsilon \sum_{j \in J_i} g_{ij}(x_i - x_j) - I_i(\epsilon) \right] \quad (4.4)$$

where K_i, J_i are index sets representing nodes connected to element i . Constant I_i , which depends on and is differentiable with respect to ϵ , is a net injection (of power or current) at node i . Its dependence on ϵ will be discussed later. A fundamental property of weakly connected networks is that neglecting the weak connection terms ϵg_{ij} results in ν isolated "areas" $\alpha=1, \dots, \nu$. Area α contains n_α connected nodes and its equation is obtained by setting $\epsilon=0$ in (4.4)

$$\frac{dx_i}{d\tau} = -\frac{1}{m_i} \left[\sum_{k \in K_i} f_{ik}(x_i - x_k) - I_i(0) \right], \quad i \in \alpha, \quad \alpha=1, \dots, \nu. \quad (4.5)$$

When the states in each area are ordered consecutively the $n \times \nu$ partition matrix U is

$$U = \text{diag} (u_1, \dots, u_v) \quad (4.6)$$

where u_a is an n_α -vector with all elements one.

Assumption 4.1 Each of the v areas formed by setting $\varepsilon=0$ in (4.4) has an equilibrium state.

In the case of power systems the above assumption requires that every area, when isolated from the rest of the system, has its own load flow. This will only be possible if the area adjusts its net injections $I_i(\varepsilon)$ so that the power exchanged with other areas is compensated internally. Thus, the injections $I_i(\varepsilon)$ are made functions of the strength of the inter-area connections ε . The choice of the dependence of I_i on ε and its impact on the accuracy of the reduced models will be discussed later in the section.

In each area α we select a reference node x_r , $r \in \alpha^{(*)}$, and form the difference $s_{ir} = x_i^e - x_r^e$ for $r \in \alpha$ and $\forall i \in \alpha$, $i \neq r$, where x_i^e, x_r^e are the values of x_i, x_r at an equilibrium x^e of the area model (4.5).

Theorem 4.2 System (4.5) has an equilibrium manifold S described by

$$\varphi_i(x) = x_i - x_r - s_{ir} = 0 \quad (4.7)$$

for $r \in \alpha$; $\forall i \in \alpha$, $i \neq r$ and all areas $\alpha=1, \dots, v$. The dynamic manifold F_{x_0} for $x(0)=x_0$ is

$$\sigma_\alpha(x) - \sigma_\alpha(x_0) = 0, \quad \alpha=1, \dots, v \quad (4.8)$$

where

$$\sigma_\alpha(x) = \sum_{i \in \alpha} m_i x_i / \sum_{i \in \alpha} m_i \quad (4.9)$$

(*)Abusing notation, we let α be the index of an area as well as the set of node indices in the area.

Furthermore,

$$y_{\alpha}(x) = \sigma_{\alpha}(x) \quad , \quad z_i = \varphi_i(x) \quad (4.10)$$

are ν slow and $\rho=n-\nu$ fast variables satisfying Theorem 3.2.

Proof: Any x satisfying

$$x_i - x_k = x_i^e - x_k^e \quad \forall i, k \in \alpha, \quad \alpha=1, \dots, \nu \quad (4.11)$$

is an equilibrium of (4.5). Following a path from node i to node r , these relations can also be written as

$$x_i - x_r = (x_i - x_{i+1}) + (x_{i+1} - x_{i+2}) + \dots + (x_{r-2} - x_{r-1}) + (x_{r-1} - x_r) = x_i^e - x_r^e \quad (4.12)$$

$r \in \alpha; \quad \forall i \in \alpha, \quad i \neq r, \quad \forall \alpha$

which is the expression in (4.7).

Writing (4.5) at an equilibrium

$$0 = - \frac{1}{m_i} \left[\sum_{k \in K_i} f_{ik}(x_i^e - x_k^e) - I_i(0) \right] \quad i \in \alpha, \quad \alpha=1, \dots, \nu \quad (4.13)$$

and using (4.1) we obtain

$$\sum_{i \in \alpha} I_i(0) = 0 \quad , \quad \alpha=1, \dots, \nu \quad (4.14)$$

The last relation gives

$$\sum_{i \in \alpha} m_i (dx_i/d\tau) = 0 \quad , \quad \alpha=1, \dots, \nu \quad (4.15)$$

where (4.1) was used once more. The dynamic manifold (4.8-4.9) is obtained by integrating and scaling (4.15). Finally the transformation (4.10) is an application of Theorem 3.2.

Note that although the model is nonlinear both the equilibrium manifold (4.7) and the dynamic manifold (4.8) are linear leading to a linear transformation separating the time scales. Manifold S is linear because the right-hand side of (4.5) is a function of a linear combination (the differences) of the states as opposed to being a function of the states individually. Manifolds F are linear because the conservation property is linear. In the case of RC-circuits the conservation property expresses Kirchhoff's current law (KCL) and in the case of power systems the conservation of angular momentum. These physical laws are linear even when some elements of the network have nonlinear characteristics.

To rewrite (4.10) in matrix form we define the difference matrix $G = \text{diag}(G_1, \dots, G_v)$ where

$$G_{\alpha} = \begin{bmatrix} -1 & 1 & 0 & . & . & . & 0 \\ -1 & 0 & 1 & . & . & . & 0 \\ . & . & . & . & . & . & . \\ -1 & 0 & . & . & . & . & 1 \end{bmatrix} \quad (4.16)$$

is a $(n_{\alpha}-1) \times n_{\alpha}$ matrix with two nonzero elements per row. Ordering the states x in the same area, consecutively, with the reference state first, denoting $M = \text{diag}(m_1, \dots, m_n)$ and $C_a = (U^T M U)^{-1} U^T M$ the transformation (4.10) is

$$\begin{bmatrix} y \\ z \end{bmatrix} = \begin{bmatrix} C_a \\ G \end{bmatrix} x - \begin{bmatrix} 0 \\ s \end{bmatrix} \quad (4.17)$$

where s is a ρ -vector with components s_{ir} . The inverse of (4.17) is

$$x = [U \quad B] \begin{bmatrix} y \\ z+s \end{bmatrix} \quad (4.18)$$

where $B = M^{-1} G^T (GM^{-1} G^T)^{-1}$. Recall from (4.6) that every row of U has one entry 1 and the rest 0, whence,

$$x_i = y_{\alpha} + b_i(z+s) \quad \forall i \in \alpha, \quad \alpha=1, \dots, v \quad (4.19)$$

where b_i is the i row of B . After simple manipulations we obtain the transformed model

$$\frac{dy_{\alpha}}{d\tau} = - \frac{1}{\sum_{i \in \alpha} m_i} [\epsilon \sum_{i \in \alpha} \sum_{\substack{j \in J_i \\ \beta=1, \dots, v \\ \beta \neq \alpha}} g_{ij} [(y_{\alpha} - y_{\beta}) + (b_i - b_j)(z+s)] - \sum_{i \in \alpha} I_i(\epsilon)] \quad (4.20)$$

$$\begin{aligned} \frac{dz_i}{d\tau} = & - \frac{1}{m_i} \left[\sum_{\substack{k \in K_i \\ k \neq r}} f_{ik} (z_i - z_k + s_{ir} - s_{kr}) + f_{ir} (z_i + s_{ir}) - I_i(\epsilon) \right] \\ & - \frac{1}{m_r} \left[\sum_{k \in K_r} f_{kr} (z_k + s_{kr}) + I_r(\epsilon) \right] \\ & - \epsilon \left[\frac{1}{m_i} \sum_{j \in J_i} g_{ij} [(y_{\alpha} - y_{\beta}) + (b_i - b_j)(z+s)] \right. \\ & \left. - \frac{1}{m_r} \sum_{r \in J} g_{rj} [(y_{\alpha} - y_{\beta}) + (b_r - b_j)(z+s)] \right] \end{aligned} \quad (4.21)$$

In (4.20), (4.21), $i \in \alpha, j \in \beta$ and b_i, b_j, b_r are the i, j, r rows of B , respectively, $r \in \alpha$. Since from (4.14) $\sum I_i(0) = 0, \alpha=1, \dots, v$

$$\sum_{i \in \alpha} I_i(\varepsilon) = 0(\varepsilon) \quad , \quad \alpha=1, \dots, \nu \quad (4.22)$$

and the right-hand side of (4.20) is $0(\varepsilon)$.

In the original state description (4.4) areas cannot be considered weakly coupled because over a longer period their interaction through weak connections becomes significant. In the transformed description the fast-time area models in z variables are weakly coupled, and the long term area interaction is approximately described by (4.20), that is, the aggregate variables y alone. The fundamental difference between this model and the original is that the decoupled z -equations obtained by setting $\varepsilon=0$ in (4.21), no longer have a continuum of equilibrium points.

The definition of slow coherency as given in [24-28] is based on a modal decomposition and is not directly applicable to nonlinear systems. Since we have shown that the two-time-scale properties remain valid for nonlinear systems, we will use them for the following generalization of the notion of slow coherency.

Slow coherency. States x_i, x_j of (4.4) are said to be slow coherent if $x(0) \in S$ implies $x_i(t) - x_j(t) = \text{const} \ \forall t \geq 0$. States x_i, x_j are said to be near slow coherent if there exists a bounded function of time $\zeta(t)$ such that

$$x(0) \in S \Rightarrow x_i(t) - x_j(t) = \text{const.} + \varepsilon \zeta(t), \ \forall t \geq 0. \quad (4.23)$$

An area is slow coherent if any two states in the area are near slow-coherent. The following theorem relates weakly coupled areas and slow coherent areas extending the corresponding result in [24-28].

Theorem 4.3 If (4.21) satisfies Assumption 1.2 and ϵ is sufficiently small, system (4.4) has ν slow-coherent areas specified by U .

Proof: From Theorem 4.2, if $i, j \in \alpha$

$$x_i - x_j = (x_i - x_r) - (x_j - x_r) = z_i - z_j + s_{ir} - s_{jr}. \quad (4.24)$$

If $x(0) \in S$, $z(0) = \varphi(x(0)) = 0$, which combined with (3.16) and (1.4) implies that $z(t) = 0(\epsilon)$. Then (4.23) follows from (4.24).

Model (4.20), (4.21) is in the explicit form (1.1). Hence, letting $\epsilon \rightarrow 0$, $I_\alpha = \lim_{\epsilon \rightarrow 0} \frac{\sum I_i(\epsilon)}{\epsilon}$ and using the fact (Theorem 3.2, Theorem 4.2) that $\bar{z} = 0$ the slow model is

$$\begin{aligned} dy_\alpha / dt = & -(1/\sum_{i \in \alpha} \sum_{\substack{j \in J_i \\ \beta=1, \dots, \nu \\ \beta \neq \alpha}} g_{ij} [(\bar{y}_\alpha - \bar{y}_\beta) + (b_i - b_j)s] \\ & - \sum_{i \in \alpha} \hat{I}_i] \end{aligned} \quad \alpha=1, \dots, \nu. \quad (4.25)$$

The fast model is

$$\begin{aligned} d\tilde{z}_i / d\tau = & -(1/m_i) [\sum_{\substack{k \in K_i \\ k \neq r}} f_{ik} (\tilde{z}_i - \tilde{z}_k + s_{ir} - s_{kr}) \\ & + f_{ir} (\tilde{z}_i + s_{ir}) - I_i(0)] \\ & - (1/m_r) [\sum_{k \in K_r} f_{kr} (\tilde{z}_k + s_{kr}) + I_r(0)] \\ & i \in \alpha, \quad i \neq r \quad \alpha=1, \dots, \nu. \end{aligned} \quad (4.26)$$

Slow model (4.25) represents an aggregate dynamic network with storage elements

$$m_{\alpha} = \sum_{i \in \alpha} m_i \quad \alpha=1, \dots, v \quad (4.27)$$

net injections

$$I_{\alpha} = \sum_{i \in \alpha} \hat{I}_i \quad \alpha=1, \dots, v \quad (4.28)$$

and interconnection characteristics

$$G_{\alpha\beta} (\bar{y}_{\alpha} - \bar{y}_{\beta}) = \sum_{\substack{i \in \alpha \\ j \in \beta}} g_{ij} [\bar{y}_{\alpha} - \bar{y}_{\beta} + (b_i - b_j)s]. \quad (4.29)$$

The aggregate model (4.25) is decoupled from the local models (4.26). Since the sums in (4.26) involve nodes from the same area, the equations for two different areas are decoupled, that is, the fast models (4.26) are local in the sense that they involve quantities from one area only. Thus, each area uses its local model and at the same time provides the data and receives results from the global model. This multimodeling decomposition helps in formulation of decentralized controls [57].

In (4.4) and in subsequent derivations it was assumed that the dependence of the injections $I_i(\varepsilon)$ on ε is known. In a realistic situation ε has a specific value and injections are constant. The dependence on ε is an asymptotic tool guaranteeing that the isolated areas formed by $\varepsilon \rightarrow 0$ have a well defined equilibrium. Therefore, for any function $I(\varepsilon)$

satisfying $\sum_{i \in \alpha} I_i(0) = 0$, $\alpha=1, \dots, v$. Assumption 4.1 holds and the quantities s_{ir} in (4.7) are well defined. However, the freedom in choosing $I(\varepsilon)$ can be utilized to influence the accuracy of reduced models in realistic systems where ε may not be very small.

Note that for $\varepsilon > 0$, the equilibria of (4.25), (4.26) are generally different from the equilibria of (4.20), (4.21). It has been observed in numerical experiments that the approximation of the time response improves when the equilibrium of the reduced models is closer to the equilibrium of the original (4.20), (4.21). It is desirable to make the two equilibria as close as possible, particularly for oscillatory responses, and if the reduced model (4.25), (4.26) is used for stability analysis. The following corollaries provide guidance in this direction.

Corollary 4.4 Let x^E be the equilibrium of (4.4). The equilibria of (4.25), (4.26) are equal to the equilibria of (4.20), (4.21) if and only if

$$s = Gx^E. \quad (4.30)$$

Proof: First note that by Theorem 3.2 and Theorem 4.2 $\tilde{z} = 0$ is the equilibrium of (4.26), irrespective of the choice of s . From (4.17) the equilibrium of (4.21) corresponding to x^E is

$$z^E = Gx^E - s \quad (4.31)$$

which is made zero by (4.30). Setting $z = 0$ in (4.20) to obtain (4.25) does not alter the equilibrium because $z^E = 0$. The choice (4.30) is unique because (4.31) is linear in s .

Corollary 4.4 shows that there is a unique s for which the equilibria of the exact and approximate systems are equal. Since by definition $s = Gx^e$ and G is a matrix of full rank, (4.30) implies that

$$x^e = x^E. \quad (4.32)$$

The next corollary gives a necessary condition on $I_i(\epsilon)$ such that (4.32) is satisfied. Boundary nodes are nodes to which interarea connections are attached.

Corollary 4.5 Equation (4.32) is satisfied only if

$$I_i(\epsilon) - I_i(0) = \epsilon \sum_{j \in J_i} g_{ij} (x_i^E - x_j^E) \quad (4.33)$$

that is, the net injection at boundary nodes is adjusted by the interarea flow while it is left unaltered at nonboundary nodes.

Proof: For the equilibrium of (4.4)

$$\sum_{k \in K_i} f_{ik} (x_i^E - x_k^E) = - \epsilon \sum_{j \in J_i} g_{ij} (x_i^E - x_j^E) + I_i(\epsilon) \quad (4.34)$$

and for the equilibrium of (4.5)

$$\sum_{k \in K_i} f_{ik} (x_i^e - x_k^e) = I_i(0). \quad (4.35)$$

Hence, (4.33) is necessary for (4.32) to be true.

In cases such as water distribution networks where some storage elements may be nonlinear, the reduction procedure is still applicable after some modification. Assuming that the stored quantity is a strictly

monotonic function of the potential, the dynamics of the network are described by an equation analogous to (4.2) in which m_i is now a function of x_i . Equation (4.9) becomes

$$\sigma_\alpha(x) = \frac{\sum_{i \in \alpha} m_i(x_i) x_i}{\sum_{i \in \alpha} m_i(x_i)} \quad (4.36)$$

and the dynamic manifold is no longer linear. The equilibrium manifold, however, is still given by (4.7).

4.3 Power System Application

The concepts of area aggregation and slow coherency that emerged from the separation of time scales in dynamic networks (Theorem 4.2) originated as model order reduction techniques in power systems [21-23]. First, a group of coherent generators, that is, generators that "swing together" is identified and then this group is replaced with an equivalent generator. Analytical studies of coherency [58,59] and coherency based aggregation [24-28] were based on linearized versions of the electromechanical model of power systems. In this section we apply the reasoning of Section 4.2 to extend the model simplification approach in [24-28] to the nonlinear electromechanical model, and to more complex models involving flux decay dynamics and voltage regulator.

The well known electromechanical model [60] of multimachine power systems is

$$2H_i \ddot{\delta}_i = P_{mi} - P_{ei} \quad i=1,2,\dots,n \quad (4.37)$$

where δ_i is the rotor angle of machine i , H_i is its inertia constant, P_{mi} and P_{ei} are its mechanical input power and electrical output power, respectively, and the small damping was ignored [25]. In this model P_{mi} is assumed to be constant and P_{ei} is given by

$$P_{ei} = \sum_{\substack{j=1 \\ j \neq i}}^n v_i v_j B_{ij} \sin(\delta_i - \delta_j) + v_i^2 G_{ii} \quad (4.38)$$

where v_i is the constant voltage "behind the transient reactance," B_{ij} is the (ij) -th entry of the admittance matrix reduced to the machine nodes and G_{ii} represents the load conductance at node i .

Substituting (4.38) into (4.37) we obtain

$$\ddot{\delta}_i = -\frac{1}{2H_i} \left[\sum_{\substack{j=1 \\ j \neq i}}^n v_i v_j B_{ij} \sin(\delta_i - \delta_j) - (P_{mi} - v_i^2 G_{ii}) \right] \quad (4.39)$$

which is in the form of (4.3) with $m_i = 2H_i$, $x_i = \delta_i$,

$$f_{ij}(x_i - x_j) = v_i v_j B_{ij} \sin(\delta_i - \delta_j) \quad (4.40)$$

$$I_i = P_{mi} - v_i^2 G_{ii} \quad (4.41)$$

Multimachine power systems are often comprised of groups of tightly connected machines with weak connections joining the groups. Assuming that weak connections are known system (4.39) takes the form (4.4) for which Theorem 4.2 gives equilibrium and dynamic manifolds and defines slow and fast variables. Note, however, that since damping was neglected the response of (4.39) is purely oscillatory and the separation of time scales

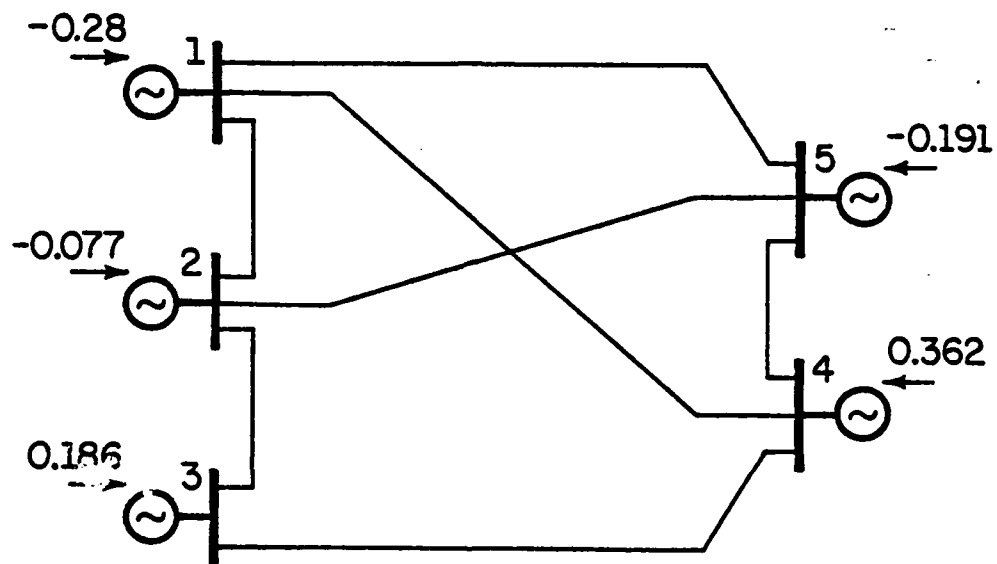
is understood in the sense of separating low frequency from high frequency oscillations [61]. Using transformation (4.10) we can arrive at expressions similar to (4.20)-(4.21) from which the slow core and the fast residues are defined as in (4.25)-(4.26). The reduction procedure in power systems and some physical interpretations of the reduced models are illustrated through the following five-machine example.

In the power system of Fig. 4.1, $H_i = 0.5$, $v_{i=1}$, $i=1, \dots, 5$ and $B_{12}=B_{23}=B_{45}=1$, $B_{34}=B_{25}=B_{15}=B_{14}=0.1$. The net injections I_i and the resulting steady-state angles (in radians) are given in columns 1 and 2 of Table 4.1.

Table 4.1. Bus angles for five-machine power example

1	2	3
$I_1 = -0.28$	$\delta_1 = 0$	$\delta_1 = 0$
$I_2 = -0.077$	$\delta_2 = 0.171$	$\delta_2 = 0.215$
$I_3 = 0.186$	$\delta_3 = 0.391$	$\delta_3 = 0.458$
$I_4 = 0.362$	$\delta_4 = 0.723$	$\delta_4 = 1.042$
$I_5 = -0.191$	$\delta_5 = 0.456$	$\delta_5 = 0.730$

Note that since admittances B_{34} , B_{25} , B_{15} , B_{14} are much smaller than the rest, the system is divided into two weakly connected areas $\alpha = \{1, 2, 3\}$, $\beta = \{4, 5\}$. Suppose now that line B_{14} is tripped and we want to simulate the resulting oscillations using reduced models (4.25)-(4.26). The post fault load flow (shown in column 3 of Table 4.1) gives



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Fig. 4.1 Five-machine power system example.

$$\begin{aligned}
s_{12} &= \delta_1^e - \delta_2^e = -0.215 \\
s_{32} &= \delta_3^e - \delta_2^e = 0.243 \\
s_{45} &= \delta_4^e - \delta_5^e = 0.312.
\end{aligned}
\tag{4.42}$$

Defining new variables

$$y_1 = \frac{\delta_1 + \delta_2 + \delta_3}{3}, \quad y_2 = \frac{\delta_4 + \delta_5}{2}
\tag{4.43}$$

$$z_1 = \delta_1 - \delta_2 - s_{12}, \quad z_2 = \delta_3 - \delta_2 - s_{32}, \quad z_3 = \delta_4 - \delta_5 - s_{45}$$

and letting $\epsilon \rightarrow 0$ we obtain the slow model

$$\begin{aligned}
\ddot{\bar{y}}_1 &= -0.033 \sin(\bar{y}_1 - \bar{y}_2 - 0.068) - 0.033 \sin(\bar{y}_1 - \bar{y}_2 + 0.147) \\
&\quad - 0.033 \sin(\bar{y}_1 - \bar{y}_2 + 0.078) - 0.057 \\
\ddot{\bar{y}}_2 &= -0.05 \sin(\bar{y}_2 - \bar{y}_1 - 0.078) - 0.05 \sin(\bar{y}_2 - \bar{y}_1 - 0.147) \\
&\quad - 0.05 \sin(\bar{y}_2 - \bar{y}_1 + 0.068) + 0.086
\end{aligned}
\tag{4.44}$$

and the fast model

$$\ddot{\tilde{z}}_1 = -2 \sin(\tilde{z}_1 - 0.215) - \sin(\tilde{z}_2 + 0.243) - 0.185
\tag{4.45}$$

$$\ddot{\tilde{z}}_2 = -2 \sin(\tilde{z}_2 + 0.243) - \sin(\tilde{z}_1 - 0.215) + 0.269$$

$$\ddot{\tilde{z}}_3 = -2 \sin(\tilde{z}_3 + 0.312) + 0.614
\tag{4.46}$$

Note that (4.44) is decoupled from (4.45), (4.46). The aggregate model (4.44) represents the oscillations of the aggregate angles y_1, y_2 against each other, whereas, the local models (4.45), (4.46) represent the

intermachine oscillations in areas α, β , respectively. Figures 4.2-4.5 show simulation curves with initial conditions equal to the prefault equilibrium. Figures 4.2, 4.3 show exact (solid lines) and approximate (dotted lines) responses of angles δ_1, δ_4 , whereas, Figures 4.4, 4.5 show exact (dotted lines) and approximate (solid lines) responses of the transformed variables y_1, z_2 . Note that generator angles are mixed variables, whereas, y_1 is predominantly slow and z_2 is predominantly fast.

We now turn to more complex models of power systems and again investigate the effect of weak connections on the time scale behavior of the system. The model we employ is basically the one in [62] with a slight simplification; we do not include the fictitious quadrature axis coil g which is meant to model eddy currents in the rotor. With this simplification the model is

$$\dot{\delta}_i = 377(\omega_i - 1) \quad (4.47a)$$

$$2H_i \dot{\omega}_i = \frac{P_{mi}}{\omega_i} - e'_{qi} i_{qi} - D_i(\omega_i - 1) \quad (4.47b)$$

$$\dot{e}'_{qi} = \frac{1}{T'_{doi}} [e'_{qi} - (x_{di} - x'_{di}) i_{di} + E_{fdi}] \quad (4.47c)$$

$$\dot{E}_{fdi} = \frac{1}{T_{Fi}} (-R_{Fi} + \frac{K_F}{T_F} E_{fdi}) \quad (4.47d)$$

$$\dot{E}_{fdi} = \frac{1}{T_{Ei}} [-(K_E + S_E (E_{fdi})) E_{fdi} + V_{Ri}] \quad (4.47e)$$

$$i_{di} = - \sum_j B_{ij} e'_{qj} \cos (\delta_i - \delta_j) \quad (4.48a)$$

$$i_{qi} = \sum_j B_{ij} e'_{qj} \sin (\delta_i - \delta_j) \quad (4.48b)$$

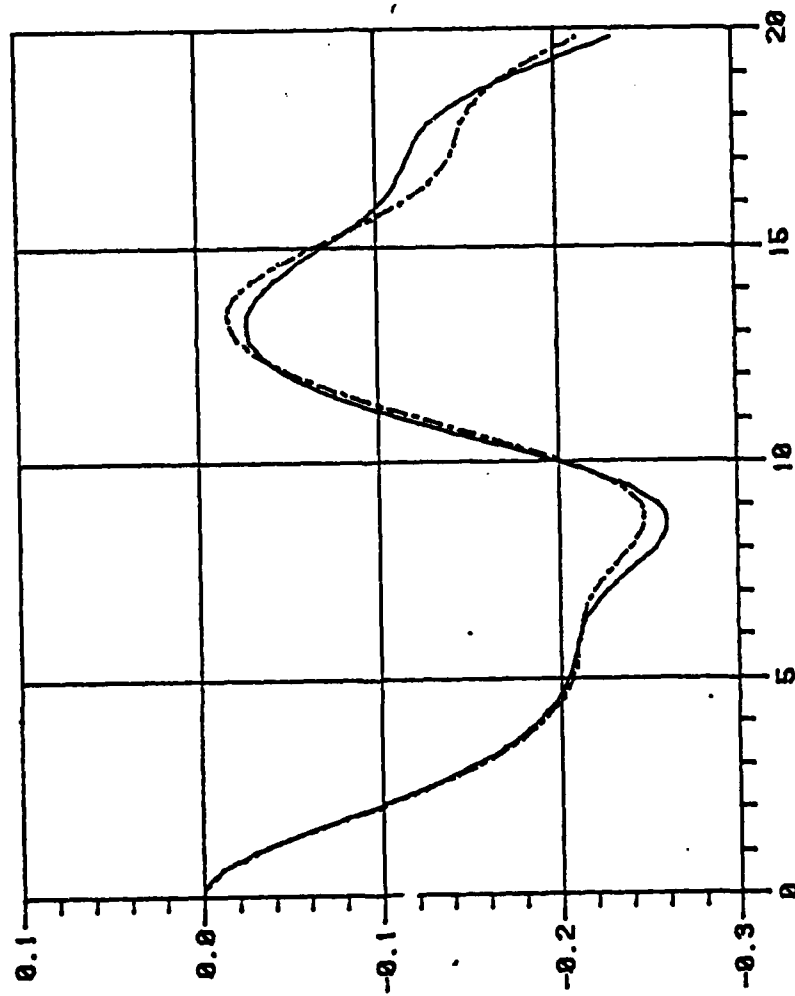


Fig. 4.2 Exact (solid line) and approximate (dotted line) response of δ_1 .

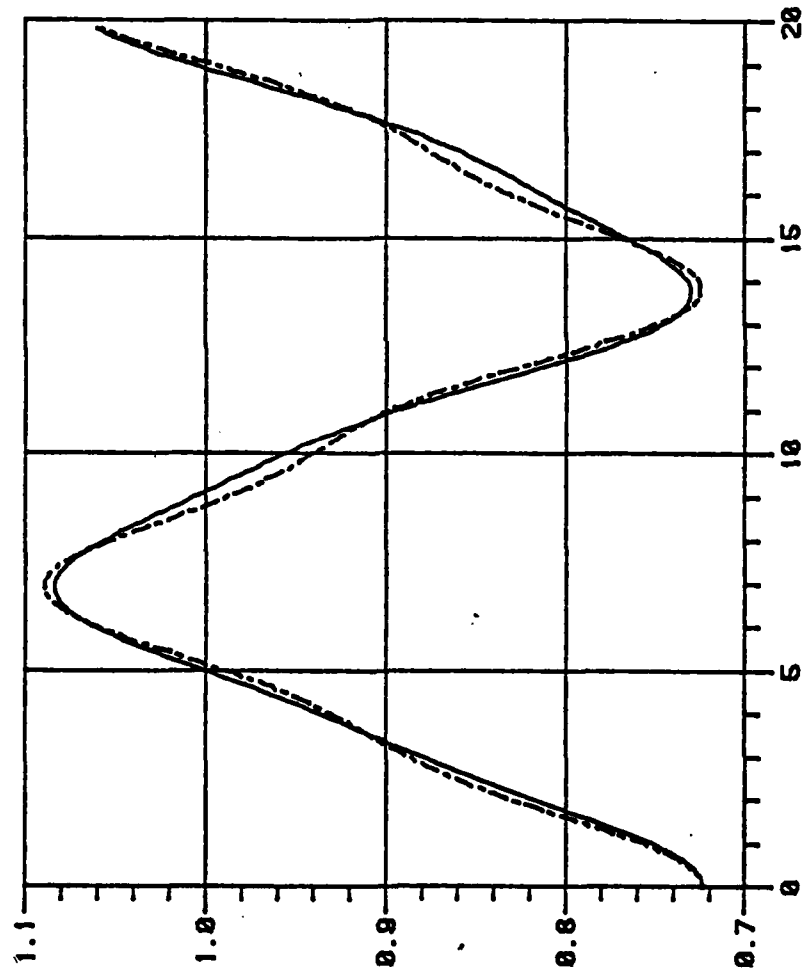


Fig. 4.3 Exact (solid line) and approximate (dotted line) responses of δ_4 .

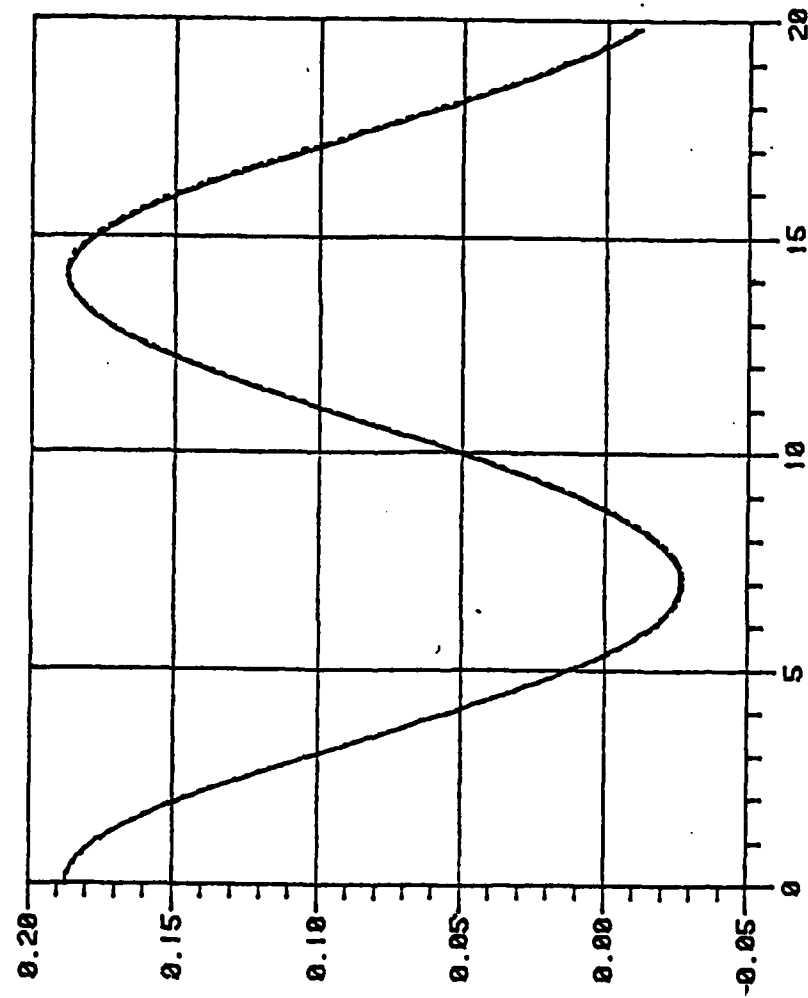


Fig. 4.4 Exact (dotted line) and approximate (solid line) responses of y_1 .

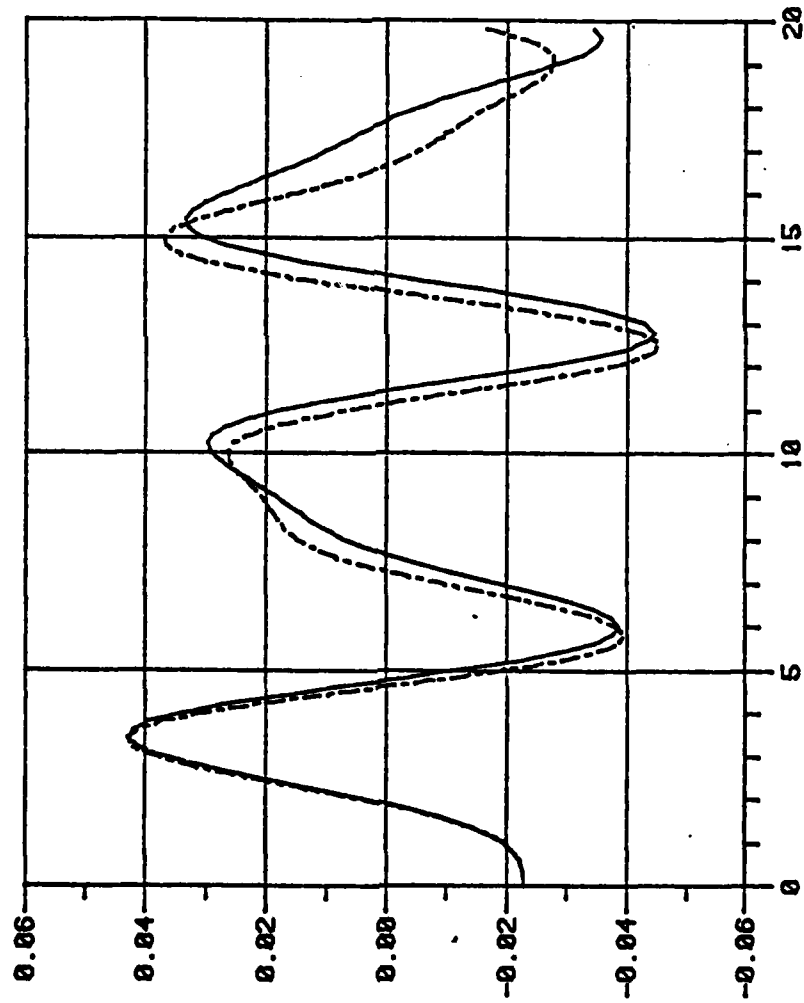


Fig. 4.5 Exact (dotted line) and approximate (solid line) responses of z_2 .

$$V_{ti} = \sqrt{V_{di}^2 + V_{qi}^2}, \quad V_{di} = e'_{di} + x'_{qi} i_{qi}, \quad V_{qi} = e'_{qi} - x'_{di} i_{di} \quad (4.49)$$

where state e'_q is proportional to the field flux, equations (4.47d-f) model the voltage regulator-exciter system and (4.48), (4.49) give the interaction of the generators through the transmission network.

If the power system is made of ν weakly connected areas, (4.48a), (4.48b) are written as

$$i_{di} = - \sum_{j \in \alpha} B_{ij} e'_{qj} \cos(\delta_i - \delta_j) - \epsilon \sum_{k \notin \alpha} B_{ik} e'_{qk} \cos(\delta_i - \delta_k) \quad (4.50a)$$

$$i_{qi} = \sum_{j \in \alpha} B_{ij} e'_{qj} \sin(\delta_i - \delta_j) + \epsilon \sum_{k \notin \alpha} B_{ik} e'_{qk} \sin(\delta_i - \delta_k) \quad (4.50b)$$

where $i \in \alpha$, $\alpha=1, \dots, \nu$. We now make the important observation that (i) machines interact solely through currents i_d, i_q and (ii) i_d, i_q are functions of the differences $\delta_i - \delta_j$ of angles, not of angles individually. Hence, letting $\epsilon=0$ in (4.50) and setting the right-hand side of (4.47) equal to zero we see that if an assumption analogous to 4.1 is met, points satisfying

$$\begin{aligned} \delta_i - \delta_r - s_{ir} &= 0 \\ \omega_i &= 1 \\ e'_{qi} &= e'_{qie} \\ R_{fi} &= R_{fie} \\ E_{fdi} &= E_{fdie} \\ V_{Ri} &= V_{Rie} \end{aligned} \quad (4.51)$$

where subscript e denotes equilibrium and $r \in \alpha$, $i \in \alpha$, $i \neq r$, $\alpha=1, \dots, \nu$ are equilibrium points of (4.47). That is, (4.47) at $\epsilon=0$ has a ν -dimensional equilibrium manifold described by (4.51). An argument similar to the one in Theorem 4.2 shows that

$$\sigma(\delta, \omega) = \sum_{i \in \alpha} \frac{D_i}{377} \delta_i + \sum_{i \in \alpha} 2H_i \omega_i = \sigma(\delta(0), \omega(0)) \quad (4.52)$$

for $\alpha=1, \dots, \nu$ defines the family of dynamic manifolds of (4.47). In all realistic cases $D_i \ll 377$ so that we can ignore the first term in (4.52). Note that although we started with a higher-dimensional model of the power system the equilibrium and conservation relations (4.51), (4.52) involving δ, ω variables are identical to the ones obtained by working with the electromechanical model. Consequently, we obtain an electromechanical slow model involving the aggregate variables $\delta_\alpha = (\sum_{i \in \alpha} 2H_i \delta_i) / \sum_{i \in \alpha} 2H_i$, $\omega_\alpha = \dot{\delta}_\alpha$.

4.4 Coherency and Localizability

As shown in Section 4.2 weak connections in a dynamic network give rise to slow-coherent groups of states which are described by local models. We now restrict ourselves to linear time invariant systems and investigate the relation between coherency and localizability when weak connections are not present. This discussion clarifies the presentation in [24,25] where the two notions were essentially treated as equivalent. For the sake of completeness we repeat here the definition of coherency of [24,25].

Definition 4.6 Let $\dot{x}=Ax$ be a LTI system, σ be a subspectrum of A and V_σ be the corresponding eigenspace. Then states x_i and x_j are said to be coherent with respect to σ if and only if, $x(0) \in V_\sigma$ implies that

$$x_i(t) = x_j(t) \quad , \quad \forall t \geq 0 \quad . \quad (4.53)$$

A group of states is said to be a σ -coherent group if any two states from the group are coherent with respect to σ .

Suppose now that n_α states of a system are considered to belong to a group α . The criterion for such grouping can be geographic proximity, accessibility to remote sensing or similar. In an attempt to describe the "local" behavior in the group we use $n_\alpha - 1$ differences $x_i - x_k$, where i, k belong to the α set of indices. Typically, we fix index k as the local reference and take all $i \neq k$ in the group α to form the differences. We then investigate under what conditions the local variables

$$z = \begin{bmatrix} G & | & 0 \end{bmatrix} x = G_\alpha x \quad (4.54)$$

where

$$G = \begin{matrix} \xleftarrow{n_\alpha} & & \xrightarrow{\hspace{1.5cm}} \\ \begin{bmatrix} -1 & 1 & 0 & . & . & . & 0 \\ -1 & 0 & 1 & . & . & . & 0 \\ . & . & . & & & & . \\ -1 & 0 & . & . & . & . & 1 \end{bmatrix} & \begin{matrix} \uparrow \\ n_\alpha - 1 \\ \downarrow \end{matrix} \end{matrix} \quad (4.55)$$

are independent of the rest of the system.

Definition 4.7 Group α is said to be localizable if there is an A_α such that (*)

$$\dot{z} = A_\alpha z \quad (4.56)$$

Lemma 4.8 A group of states is localizable if and only if it is a σ -coherent group and

$$n_\sigma = n - n_\alpha + 1 \quad (4.57)$$

where n_σ is the number of modes in σ .

Proof: If the group is localizable z can be decoupled from system implying that $n - n_\alpha + 1$ modes are unobservable from z . If V_σ is a basis of the eigenspace corresponding to these modes

$$G_\sigma V_\sigma = 0 \quad (4.58)$$

which implies that rows of V_σ corresponding to states in the group are equal. Hence, it is a coherent group. Conversely, the rows of V_σ corresponding to a coherent group are equal implying that n_σ modes are unobservable. When n_σ satisfies (4.57) the number of observable modes is $n_\alpha - 1$ which equals the dimension of z and the group is localizable.

Note that for V_σ to be full rank n_σ has to satisfy

$$n_\sigma \leq n - n_\alpha + 1 \quad (4.59)$$

(*) The notion of localizability is identical to aggregability with respect to matrix G_σ of (4.54). To avoid confusion we reserve the latter term for "area" aggregation.

Hence, for the coherent group to be localizable n_σ is required to take its maximum permissible value. Note also that the smaller the group the larger n_σ and the harder it is to satisfy the localizability conditions. On the other hand we do not benefit much by localizing a large group. Finally, note that the only modes observable from the local variables of a localizable group are the complementary modes σ^c , henceforth, called local modes.

When a system is divided into more groups of states the localizability conditions can be applied independently to each group. As an example, let the modal matrix of a 5-state system be

$$\begin{array}{ccccc}
 & \lambda_1 & \lambda_2 & \lambda_3 & \lambda_4 & \lambda_5 \\
 \begin{array}{l} x_1 \\ x_2 \\ x_3 \\ x_4 \\ x_5 \end{array} & \left[\begin{array}{ccccc} a & b & c & * & * \\ a & b & c & * & * \\ a & b & c & * & * \\ * & n & k & l & m \\ * & n & k & l & m \end{array} \right] & \left. \begin{array}{l} \\ \\ \\ \end{array} \right\} \begin{array}{l} \alpha \\ \\ \beta \end{array}
 \end{array} \quad (4.60)$$

where the stars can be any numbers such that the matrix is nonsingular.

Group $\alpha = \{x_1, x_2, x_3\}$ is a σ_α -coherent group where $\sigma_\alpha = \{\lambda_1, \lambda_2, \lambda_3\}$ and satisfies (4.57). Hence, it is localizable and its local modes are $\sigma_\alpha^c = \{\lambda_4, \lambda_5\}$. Likewise group $\beta = \{x_4, x_5\}$ is $\sigma_\beta = \{\lambda_2, \lambda_3, \lambda_4, \lambda_5\}$ satisfies (4.57) and it is localizable with local modes $\sigma_\beta^c = \{\lambda_1\}$. When, as in the above example both the groups and the local modes are disjoint, the system is called multi-localizable.

As it is clear from (4.60) the multi-localizability conditions are very stringent. Note that in a multi-localizable system, each set of local variables decouples from the rest of the system and observes only the local modes. A less stringent requirement is that the local variables from the different groups decouple from the system as a single set.

Definition 4.9 Let the states of $\dot{x} = A_x x$ be divided into r disjoint groups, each consisting of two or more states, and s states not assigned to any group. Then the system is called decomposable if the local variables

$$z = \left[\begin{array}{ccc|c} G_1 & & & \\ & 0 & & \\ & & \ddots & \\ & & & G_r \\ 0 & & & \end{array} \right] \quad x = G_T x \quad (4.61)$$

decouple from the system that is if there is a matrix A_ℓ such that

$$\dot{z} = A_\ell z \quad (4.62)$$

where

$$G_v = \begin{array}{c} \updownarrow \\ n_v - 1 \end{array} \left[\begin{array}{cccccc} -1 & 1 & 0 & . & . & 0 \\ -1 & 0 & 1 & . & . & 0 \\ . & . & . & . & . & . \\ . & . & . & . & . & . \\ -1 & 0 & . & . & . & 1 \end{array} \right] \quad (4.63)$$

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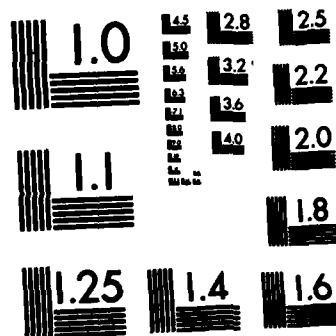
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η_v is the number of states in group v and the s single states are the last entries of x .

The following lemma establishes the relationship between decomposability and coherency.

Lemma 4.10 A system is decomposable if and only if each of its groups is coherent with respect to the same set of modes σ_a and

$$r + s = n_{\sigma_a} \quad (4.64)$$

where n_{σ_a} the number of modes in σ_a .

Proof: If a system is decomposable, only $n-s-r$ modes are observable from the same number of local variables z . Let σ_a be the set of $r+s$ unobservable modes and V a basis for the corresponding eigenspace. Then

$$G_T V = 0 \quad (4.65)$$

which implies that each of the r groups is a coherent group with respect to σ_a . Conversely if all the groups are coherent with respect to σ_a modes are unobservable from z . If further the number of the z variables $n-s-r$ equals the number of observable modes $n-n_{\sigma_a}$, that is (4.64) is satisfied, the system is decomposable.

As an illustration consider again groups α and β in (4.60).

Both α and β are coherent with respect to $\sigma_a = \{\lambda_2, \lambda_3\} = \sigma_\alpha \cap \sigma_\beta$ and the two modes σ_a equal in number the two areas. Hence, the system in (4.60) is decomposable.

CHAPTER 5

SUGGESTIONS AND CONCLUSIONS

5.1 Suggestions for Further Research

The ideas in this thesis can be extended in several directions. In Chapters 2 and 3 we showed that equilibrium and conservation properties of an auxiliary system imply multi-time-scale behavior. However, we did not investigate the relation between the two properties. Does the existence of one property imply the other? And under what conditions? It is clear that in Linear Time Invariant systems with simple structure of $\mathcal{N}(A_0)$ (Equations (2.11)-(2.13)) the two properties are equivalent. We feel that the existence of an equilibrium manifold implies conservation properties in a wide class of nonlinear systems. This issue and the one of systematic procedures for finding equilibrium and dynamic manifolds deserve further investigation. A look at the decomposition in [63] and the differential geometry techniques used therein should prove useful. Time scales in discrete-time systems is a rather neglected topic. Does the coordinate-free characterization carry over to this class of systems? And how are time scales related to the sampling period?

We have dealt mostly with time scales of free systems. On the other hand, high gain control is known to change the time-scale behavior (Section 3.3) of systems. In Section 2.5 we have given conditions under which nonexplicit controlled LTI models can be transformed to explicit controlled models. Similar results for nonlinear systems would be desirable.

This issue is related to extension of the high gain results of Section 3.3 to wider classes of systems. A rather easy extension would be to include dependence of matrix B on x .

The decomposition into a slow core and fast residues seems promising in decentralized and hierarchical control design along the lines of [57]. When dealing with physical systems such as power systems, prospects for implementation of such designs should be a consideration.

In terms of practical significance the time scale decomposition of dynamic networks in Sections 4.2, 4.3 seems to be the most promising. Stability tests by decomposition methods have been used in power systems [64] but they usually give conservative results. We feel that the decomposition into slow core and fast residues takes advantage of the structure of the system (weakly connected areas) and it is likely to provide practical results. Moreover, it can furnish information on the type of instability, that is intermachine or interarea instability. It would also be interesting to study time-scale separation and stability questions using more complex generator models and an unreduced network.

5.2 Conclusions

Singular perturbations have been related to equilibrium and conservation properties of an auxiliary system. Besides providing a coordinate-free characterization of singularly perturbed systems, these properties have been used in definition of new predominantly slow and predominantly fast coordinates. In the new coordinates an extensive amount of literature provides simplified models, asymptotic calculations, two-stage designs and stability tests.

Results on high gain feedback control have been extended to a class of systems much larger than LTI systems. Disturbance rejection behavior has been shown to be similar to the one in the LTI case.

The relation between weak connections and time scales in a class of interconnected systems has been established. Weak connections combined with equilibrium and dynamic manifolds of the subsystems give rise to multi-time-scale behavior. Separation of the time scales results in a slow core describing the system-wide dynamics and a set of fast residues describing the local dynamics. In the new representation recent stability results can be applied to give decentralized stability tests.

When the interconnected system has the added structure of a dynamic network the slow core and the fast residues acquire physical significance and the definition of slow and fast variables is related to physical laws. The linearity of these laws makes the transformation separating the time scales linear. This transformation is the area aggregation-slow coherency one, developed for linearized models of power systems.

APPENDIX I. SINGULAR PERTURBATION ON THE INFINITE HORIZON

Under consideration is the system

$$(P) \quad \begin{aligned} \frac{dy}{dt} &= f(t, y, z, \varepsilon) & y(t_0) &= y_0 \\ \varepsilon \frac{dz}{dt} &= g(t, y, z, \varepsilon) & z(t_0) &= z_0 \end{aligned}$$

with degenerate system

$$(D) \quad \frac{dy}{dt} = f(t, y, 0, 0), \quad y(t_0) = y_0$$

and boundary layer system

$$(BL) \quad \frac{dz}{d\tau} = g(\alpha, \beta, z, 0)$$

where (α, β) are treated as parameters. In (P) x, f are k -vectors, g, y are j -vectors and, without loss of generality it is assumed that $g(t, y, 0, 0) = 0$ for all t, y .

Let $|x| = \sum |x_i|$ be the norm of x , let $I = [0, \infty]$, $S_R = \{(y, z) \in E^{k+j} : |y| + |z| \leq R\}$ and let $S_R|_y, S_R|_z$ represent the restrictions of S_R to E^k and E^j .

The following assumptions are made about (P), (D), (BL).

(I) System (P) has a solution $y = y(t)$, $z = z(t)$ that exists for $t_0 \leq t < \infty$.

(II) $f, g, f_y, f_z, g_t, g_y, g_z \in C$ where f_x denotes the matrix $\frac{\partial f_i}{\partial y_j}$, $i, j = 1, \dots, k$.

(III) Function f is continuous at $z=0, \varepsilon=0$ uniformly in $(t, y) \in I \times S_R|_y$ and $f(t, y, 0, 0)$, $f_y(t, y, 0, 0)$ are bounded on $I \times S_R|_y$.

(IV) Function g is continuous at $\varepsilon=0$ uniformly in $(t, y, z) \in I \times S_R$, and $g(t, y, z, 0)$ and its derivatives with respect to t and the components of y, z are bounded on $I \times S_R$.

To simplify notation let \mathcal{X} be the class of all continuous, strictly increasing, real valued functions $d(r)$, $0 \leq r$ with $d(0) = 0$; and let \mathcal{J} be the class of all nonnegative, strictly decreasing, continuous, real-valued functions $\sigma(s)$, $0 \leq s < \infty$ for which $\sigma(s) \rightarrow 0$ as $s \rightarrow \infty$.

(V) The zero solution of D is uniform-asymptotically stable. That is, if $x = \xi(t, t_0, y_0)$ is the solution of (D) , $\exists d \in \mathcal{X}$ such that

$$|\xi(t, t_0, y_0)| \leq d(|y_0|) \sigma(t - t_0) \text{ for } |y_0| \leq R, 0 \leq t_0 \leq t < \infty.$$

(VI) The zero solution of (BL) is uniform-asymptotically stable uniformly in the parameter $(\alpha, \beta) \in I \times S_R|_y$. That is, if $y = \Psi(s, z_0, \alpha, \beta)$ is the solution of BL , $\exists e \in \mathcal{X}, \rho \in \mathcal{J}$, such that

$$|\Psi(s, z_0, \alpha, \beta)| \leq e(|z_0|) \rho(s)$$

for all $0 \leq s < \infty$, $|z_0| \leq R$ and $(\alpha, \beta) \in I \times S_R|_y$.

Then the following Theorem is true [36].

Theorem [36] Let conditions (I) through (VI) be satisfied. Then for sufficiently small $|y_0| + |z_0|$ and ε the solution of the perturbed system (P) exists for $t_0 \leq t < \infty$, and this solution converges to the solution of the degenerate system (D) as $\varepsilon \rightarrow 0^+$ uniformly on all closed subsets of $t_0 < t < \infty$.

APPENDIX II. A STABILITY THEOREM

Consider

$$(P) \quad \begin{aligned} \dot{y} &= f(y, z, \varepsilon) & y \in B_y \subset \mathbb{R}^n \\ \varepsilon \dot{z} &= g(y, z, \varepsilon) & z \in B_z \subset \mathbb{R}^m \end{aligned}$$

where B_y, B_z denote closed spheres centered around $y=0, z=0$. Assume that $y=0, z=0$ is the unique equilibrium of (P) in B_y, B_z and that $g(y, 0, 0) = 0$, for all $y \in B_y$. The reduced system of (P) is

$$(R) \quad \dot{y} = f(y, 0, 0) \triangleq f_r(y)$$

and its boundary layer is

$$(BL) \quad \frac{dz}{d\tau} = g(y, z(\tau), 0).$$

Let the following assumptions be satisfied.

(I) Reduced system (R) has a Lyapunov function $V: \mathbb{R}^n \rightarrow \mathbb{R}$, such that for all $y \in B_y$,

$$[\nabla_y V(y)]^T f_r(y) \leq -\alpha_1 \Psi^2(y), \quad \alpha_1 > 0$$

where $\Psi(y)$ is a scalar-value d function of y with $\Psi(0) = 0, \Psi(y) \neq 0$ if $y \neq 0$.

(II) Boundary-layer system (BL) has a Lyapunov function $W(y, z): \mathbb{R}^n \times \mathbb{R}^m \rightarrow \mathbb{R}$ such that for all $y \in B_y, z \in B_z$

$$[\nabla_z W(y,z)]^T g(y,z,0) \leq -\alpha_2 \phi^2(z) \quad \alpha_2 > 0$$

where $\phi(z)$ is scalar valued and $\phi(0)=0$, $\phi(z) \neq 0$ if $z \neq 0$.

(III) The following inequalities hold for all $y \in B_y$, $z \in B_z$

$$(a) \quad [\nabla_y W(y,z)]^T f(y,z) \leq C_1 \phi^2(z) + C_2 \phi(z) \psi(y)$$

$$(b) \quad [\nabla_y V(y)]^T [f(y,z) - f(y,0)] \leq \beta_1 \psi(y) \phi(z)$$

$$(c) \quad [\nabla_z W(y,z)]^T [g(y,z,\epsilon) - g(y,z,0)] \leq \epsilon K_1 \phi^2(z) + \epsilon K_2 \psi(y) \phi(z)$$

Theorem [20] If conditions (I)-(III) are true, the origin $x=0, y=0$ is an asymptotically stable equilibrium point of (P).

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